

Localization and Estimation of Unknown Forced Inputs: A Group LASSO Approach

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Abstract—We model and study the problem of localizing a set of sparse forcing inputs for linear dynamical systems from noisy measurements when the initial state is unknown. This problem is of particular relevance to detecting forced oscillations in electric power networks. We express measurements as an additive model comprising the initial state and inputs grouped over time, both expanded in terms of the basis functions (i.e., impulse response coefficients). Using this model, with probabilistic guarantees, we recover the locations and simultaneously estimate the initial state and forcing inputs using a variant of the group LASSO (linear absolute shrinkage and selection operator) method. Specifically, we provide a tight upper bound on: (i) the probability that the group LASSO estimator wrongly identifies the source locations, and (ii) the ℓ_2 -norm of the estimation error. Our bounds explicitly depend upon the length of the measurement horizon, the noise statistics, the number of inputs and sensors, and the singular values of impulse response matrices. Our theoretical analysis is one of the first to provide a complete treatment for the group LASSO estimator for linear dynamical systems under input-to-output delay assumptions. Finally, we validate our results on synthetic models and the IEEE 68-bus, 16-machine system.

Index Terms—Forced oscillations, unknown input, group LASSO, invariant zeros, source localization, sparse estimation.

I. INTRODUCTION

Low-frequency oscillations in the electric transmission grid are indicative of the type of disturbance afflicting the system. *Natural oscillations*, with frequencies in between 0.1–0.2 Hz, are triggered by random load fluctuations and sudden network switching. In contrast, *forced oscillations* (FOs), with frequencies in between 0.1–15 Hz, result from external inputs injected by malfunctioned or compromised devices, such as power system stabilizers (PSS), generator controllers and exciters, and cyclic loads etc. [1]. FOs remain undamped for longer periods of time, and if not mitigated, they pose a greater risk to the power systems operation, potentially causing blackouts.

A popular and inexpensive method to mitigate FOs in power systems is to remove the source triggering these oscillations [1]–[3]. This amounts to accurately locating the FO sources. As installing sensors at each potential source is expensive, recent research suggests using phasor measurement unit (PMU) measurements based source localization algorithms. These algorithms range from physics-based energy approaches to completely data-driven approaches [1]; the latter, albeit their impressive performance on test cases, lack theoretical guarantees. This deficiency makes it harder to quantify the

performance and limitations of measurement-based methods on what is and is not possible.

We address the lack of guarantees of existing data-driven approaches by posing the localization problem as a regularized optimization problem—referred to as the group LASSO estimator. The regularization term imposes sparsity constraints on the number of source locations, which is often the case in many practical systems, including power systems [2], [4]. The input to our optimization problem are the noisy measurements and dynamical system matrices. It returns the source locations and estimates of unknown initial state and inputs (oscillatory or not) injected by these sources. Formally, we consider

$$\underbrace{\begin{bmatrix} \hat{\mathbf{x}}_0 \\ \hat{\mathbf{u}} \end{bmatrix}}_{\hat{\boldsymbol{\beta}}} \in \arg \min_{\mathbf{x}_0, \{\mathbf{u}_j\}_{j=1}^m} \left\| \mathbf{y} - \mathbf{O}\mathbf{x}_0 - \sum_{j=1}^m \mathbf{J}_j \mathbf{u}_j \right\|_2^2 + \lambda \sum_{j=1}^m \|\mathbf{u}_j\|_2, \quad (1)$$

where $\mathbf{u}_j = [u_j[0], \dots, u_j[N]]^T$ is a vector of inputs injected by the j^{th} source, $j \in \{1, \dots, m\}$, over a discrete time horizon $\mathcal{D} \triangleq \{0, \dots, N\}$; \mathbf{y} is the noisy batch measurements collected over \mathcal{D} from multiple sensors; \mathbf{O} and \mathbf{J}_j are the observability and forced impulse response matrices, resp; and $\lambda \geq 0$ is the tuning parameter. Let $\boldsymbol{\beta}^* = (\mathbf{x}_0^*, \mathbf{u}_1^*, \dots, \mathbf{u}_m^*)$ be the unknown ground truth and $S \triangleq \{j : \mathbf{u}_j^* \neq 0\} \subset \{1, \dots, m\}$ be the set of active sources. By sparse number of sources, we mean $|S| = m^* \ll m$. Defining $\hat{S} \triangleq \{j : \hat{\mathbf{u}}_j \neq 0\}$, where $\hat{\mathbf{u}}$ as the estimate, we show that $\hat{S} = S^*$ and $\|\hat{\boldsymbol{\beta}} - \boldsymbol{\beta}^*\|_2 \leq \epsilon$, for any $\epsilon > 0$, hold with high probability.

In the context of regression models, including linear, logistic, and functional models, a rich literature exists on quantifying the theoretical performance of the group LASSO estimator and its variants; see [5]–[7]. However, these works assume \mathbf{J}_j and \mathbf{O} to be random or to satisfy rather restrictive assumptions, either of them may not hold for \mathbf{J}_j and \mathbf{O} obtained from linear dynamical systems. Further, \mathbf{J}_j associated with the non-zero input \mathbf{u}_j^* could be rank deficient, especially if the underlying linear dynamical system is only d -delay left invertible [8]; this in turn eliminates the strict convexity property of the objective in (1). As a result, there may exist multiple optimal solutions ($\hat{\boldsymbol{\beta}}$); hence, it is not clear if \hat{S} is common for all these solutions. We address all these issues by imposing physically meaningful assumptions on \mathbf{O} and \mathbf{J}_j .

Going beyond the motivating example of forced oscillations in electric power systems, the problem setup in (1) is general and the formal results in this paper can be used to localize and

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¹A dynamical system is said to be d -delay left invertible if $u_j[k]$ can be uniquely determined from noise-less measurements $\{y[k], y[k+1], \dots, y[k+d]\}$.

reconstruct sparse inputs for a variety of practical engineering systems modeled as linear dynamical systems.

Paper Contributions: The problem we introduce in (1) is distinct from state of the art regularized based optimization methods in seeking to localize inputs and estimate initial state using sufficiently delayed measurements over a block of time. For this model, our main contributions as follows.

- 1) Our first result is in deriving sufficient conditions under which the following hold with high probability: (i) the estimation error in the ℓ_2 -sense is bounded, and (ii) the localized sources match the true sources. A key contribution is that despite the rank deficiency of model matrices, we guarantee that the group LASSO can localize the sources correctly. For rank deficient matrices \mathbf{J}_i , we provide estimation guarantees for the delayed inputs (see Section III). Our result hinges on introducing and thresholding a *mutual incoherence condition* (MIC) on the augmented \mathbf{O} and \mathbf{J}_i matrices.
- 2) The time-domain MIC condition we introduce requires computing correlations among \mathbf{O} and \mathbf{J}_i . This operation is computationally hard especially for large system dimension and estimation horizon N . To tackle this hurdle, we upper bound the time-domain MIC with a frequency-domain MIC. Interestingly, the latter MIC is a sufficient condition if we were to consider a LASSO estimator in the frequency-domain. We also establish a relationship between the performance of the proposed group LASSO approach and the absence of invariant zeros for the subsystem excited by non-zero inputs, and thresholding the frequency domain MIC.
- 3) We validate the group LASSO estimator's performance on synthetic data and the IEEE 68-bus, 16-machine system. We implement our estimator using the Alternating Direction Method Multipliers (ADMM) method [9].

Related Literature: In the context of power systems, model based-approaches, e.g., energy dissipation methods based on frequency domain data and statistical signal processing methods based on AR and ARMA models, are commonly used to localize unknown forced oscillatory inputs. Reference [10] proposes a Bayesian approach to localize sources based on the generators frequency response functions. In [11], the pseudo-inverse of a set of system transfer functions are multiplied by a vector of PMU measurements to yield an FO solution vector. In [12], the authors leverage the properties of magnitude and phase responses of transfer functions between different buses to identify possible oscillation sources. Instead, completely data-driven methods include the use of novel machine learning methods on the multivariate PMU time series data in [13] and the robust PCA method in [2], which exploits the low-rank nature of PMU data and also the sparsity of the locations.

More broadly, there is a growing research on source localization and unknown state and sparse input reconstruction in dynamical systems based on sparsity constrained optimization methods. The problem of source location in the context of attacks on inputs and sensors is studied in [14] and [15]. However, these works work with noise-free measurements and do not focus on unknown input estimation. In [16], [17], by

assuming the knowledge of inputs and using randomly sampled measurements, the authors obtained sample complexity (bounds on the number of measurements) results for reconstructing the initial state with sparsity constraints. Instead, the authors in [18] and [19], consider sparse input and non-sparse state reconstruction using batch wise noise-less measurements and sequential noisy measurements, respectively. However, these works do not address location recovery guarantees for the unknown sources and the initial state. Finally, we acknowledge works in [20], [21], where the authors used banks of input observers based residual generation methods to identify source locations in noise-free systems—albeit with strong assumptions, as they do not consider sparse inputs.

In contrast to these works, we consider a unified framework, based on a LASSO method, to jointly locate the sources, and estimate the sparse inputs along with the unknown initial state. As highlighted in several other non-sparsity based input identification methods [20], [22], [23], our results also highlight the role of invariant zeros for sparse input recovery.

Mathematical Notation: We denote the vectors and matrices are by boldface lower case and upper case letters. Denote the $d \times d$ identity matrix by \mathbf{I}_d . Denote the pseudoinverse of \mathbf{X} by \mathbf{X}^\dagger . The rangespace of \mathbf{X} is defined by $\mathcal{R}(\mathbf{X}) = \{\mathbf{X}\mathbf{z} : \mathbf{z} \in \mathbb{R}^m\}$. Given $S \subset \{1, \dots, m\}$ and $\mathbf{x} \in \mathbb{R}^m$, we write \mathbf{x}_S for the sub-vector of \mathbf{x} formed from the entries of \mathbf{x} indexed by S . Similarly, we write \mathbf{M}_S for the submatrix of \mathbf{M} formed from the columns of \mathbf{M} indexed by S . For $1 \leq p < \infty$ and the vector $\mathbf{x} = [x_1, \dots, x_m]$, denote $\|\mathbf{x}\|_p = (\sum_{i=1}^m |x_i|^p)^{1/p}$. Instead, $\|\mathbf{u}\|_\infty = \max_l |u_l|$. The $\ell_{a,b}$ -mixed-norm, with $a, b \geq 0$, of $\mathbf{z} = [\mathbf{z}_1^T, \dots, \mathbf{z}_r^T]^T$ is given by $\|\mathbf{z}\|_{a,b} = \sum_{j=1}^r \|\mathbf{z}_j\|_a^b$. By convention, $\|\mathbf{z}\|_{a,0} \triangleq \sum_{j=1}^r I(\|\mathbf{z}_j\|_a \neq 0)$, where $I(\cdot)$ is the indicator function, counts the number of non-zero vectors. For a positive integer m , we denote $[m] = \{1, \dots, m\}$.

II. PROBLEM SETUP AND PRELIMINARIES

For a sampled system, we obtain a linear relation between the batch measurements and the initial state and forced inputs. We then formulate a group LASSO optimization problem for the above model to estimate the initial state and inputs, and to locate the unknown sources.

A. Linear dynamics under sparse forced inputs

Consider the following continuous-time linear system subjected to external inputs:

$$\dot{\mathbf{x}}_c(t) = \mathbf{A}_c \mathbf{x}_c(t) + \mathbf{B}_c \mathbf{u}_c^*(t), \quad t \in \mathbb{R}, \quad (2)$$

where $\Delta \mathbf{x}_c(t) \in \mathbb{R}^n$ and $\mathbf{u}_c^*(t) \in \mathbb{R}^m$ is the state and input. We assume the input to be sparse, that is $\|\mathbf{u}_c^*(t)\|_0 \leq m^* \ll m$ for all $t \in \mathbb{R}$. In the context of power systems, the state $\Delta \mathbf{x}_c(t)$ consists of the dynamical states of generators and their control systems, including rotor angles, speed deviations, field excitation voltage, etc. Instead, $\mathbf{u}_c^*(t) = [u_{c,1}^*(t), \dots, u_{c,m}^*(t)]^T$ is the vector of inputs triggered by the sources of FOs, among which only m^* locations are active. However, our model in (2), except for sparsity constraints, is general and allows for multi-dimensional un-modeled exogenous stochastic or deterministic disturbances, benign faults, or adversarial attacks.

We consider the discrete-time dynamics of (2) together with a measurement equation:

$$\mathbf{x}[k+1] = \mathbf{A}\mathbf{x}[k] + \mathbf{B}\mathbf{u}^*[k], \quad (3)$$

$$\mathbf{y}[k] = \mathbf{C}\mathbf{x}[k] + \mathbf{v}[k], \quad k = 0, 1, \dots, \quad (4)$$

where $\mathbf{A} = e^{\mathbf{A}_c \delta t}$, $\mathbf{B} = (\int_0^{\delta t} e^{\mathbf{A}_c \tau} d\tau) \mathbf{B}_c$, and δt is the sampling time period, and $\mathbf{u}^*[k] = [u_1[k], \dots, u_m[k]]^T$. Further, $\mathbf{y}[k] = [y_1[k], \dots, y_p[k]]^T \in \mathbb{R}^p$ is the measurement, $\mathbf{v}[k] \stackrel{\text{iid}}{\sim} \mathcal{N}(\mathbf{0}, \sigma^2 \mathbf{I})$ is noise, and $\mathbf{C} \in \mathbb{R}^{p \times n}$ is the sensor matrix. In Section IV, we consider dynamics in (3) with process noise, and also relax the diagonal covariance assumption on $\mathbf{v}[k]$.

Let $S = \{j : u_j[k] \neq 0 \text{ for at least one } k \geq 0\} \subset [m]$ and $S^c = [m] \setminus S$. We refer S and S^c to as the active and inactive set. Partition \mathbf{B} as $\mathbf{B} = [\mathbf{B}_S \mathbf{B}_{S^c}]$ and $\mathbf{u}[k] = [\mathbf{u}_S^T[k] \mathbf{u}_{S^c}^T[k]]^T$, with $\mathbf{u}_{S^c}^*[k] = [u_{i_1}^*[k], \dots, u_{i_r}^*[k]]$ and $\mathbf{B}_{S^c} = [\mathbf{b}_{i_1}, \dots, \mathbf{b}_{i_r}]$, where $i_r \in S^c$ and $r = |S^c| = m - m^*$. Similarly, define $\mathbf{u}_S^*[k]$ and \mathbf{B}_S . Then, the input term in (3) can be written as

$$\begin{aligned} \mathbf{B}\mathbf{u}^*[k] &= \sum_{j=1}^m \mathbf{b}_j u_j^*[k] = \sum_{j \in S} \mathbf{b}_j u_j^*[k] + \sum_{j \in S^c} \mathbf{b}_j u_j^*[k] \\ &= \mathbf{B}_S \mathbf{u}_S^*[k] + \mathbf{B}_{S^c} \mathbf{u}_{S^c}^*[k]. \end{aligned} \quad (5)$$

The above representations will play a key role in formulating our group LASSO problem in Section II-B

Using (3)-(4), we express the batch measurements \mathbf{y} (see below) as a linear model with added noise. Define the vectors

$$\mathbf{y} = \begin{bmatrix} \mathbf{y}[0] \\ \vdots \\ \mathbf{y}[N] \end{bmatrix}, \quad \mathbf{v} = \begin{bmatrix} \mathbf{v}[0] \\ \vdots \\ \mathbf{v}[N] \end{bmatrix}, \quad \text{and} \quad \mathbf{u}_j^* = \begin{bmatrix} u_j^*[0] \\ \vdots \\ u_j^*[N] \end{bmatrix}, \quad (6)$$

where $\mathbf{y}, \mathbf{v} \in \mathbb{R}^{p(N+1)}$ and $\mathbf{u}_j^* \in \mathbb{R}^{N+1}$, for all $j \in S^c$. Here, $N+1$, with $N > 0$ is the length of the estimation horizon. We also define the observability matrix $\mathbf{O} \in \mathbb{R}^{p(N+1) \times n}$ and the impulse response matrix $\mathbf{J}_j \in \mathbb{R}^{p(N+1) \times N+1}$ as

$$\mathbf{O} = \begin{bmatrix} \mathbf{C} \\ \mathbf{C}\mathbf{A} \\ \mathbf{C}\mathbf{A}^2 \\ \vdots \\ \mathbf{C}\mathbf{A}^N \end{bmatrix}; \quad \mathbf{J}_j = \begin{bmatrix} \mathbf{H}_0^{(j)} & \mathbf{0} & \mathbf{0} & \dots & \mathbf{0} \\ \mathbf{H}_1^{(j)} & \mathbf{H}_0^{(j)} & \mathbf{0} & \dots & \mathbf{0} \\ \mathbf{H}_2^{(j)} & \mathbf{H}_1^{(j)} & \mathbf{H}_0^{(j)} & \dots & \mathbf{0} \\ \vdots & \vdots & \ddots & \ddots & \vdots \\ \mathbf{H}_N^{(j)} & \mathbf{H}_{N-1}^{(j)} & \dots & \mathbf{H}_1^{(j)} & \mathbf{H}_0^{(j)} \end{bmatrix}, \quad (7)$$

where $j \in S \cup S^c$, and the l -th impulse response (Markov) parameter, $\mathbf{H}_l^{(j)} \in \mathbb{R}^{p \times 1}$, at the j -th location is defined as

$$\mathbf{H}_l^{(j)} := \begin{cases} \mathbf{0} & \text{if } l = 0, \\ \mathbf{C}\mathbf{A}^{l-1} \mathbf{b}_j & \text{if } l \geq 1. \end{cases} \quad (8)$$

Let $\mathbf{x}[0] = \mathbf{x}_0^*$ be the unknown initial state. From (3)-(4) and the fact that $\mathbf{B}\mathbf{u}^*[k] = \sum_{j=1}^m \mathbf{b}_j u_j^*[k]$, we observe that

$$\mathbf{y} = \mathbf{O}\mathbf{x}_0^* + \sum_{j=1}^m \mathbf{J}_j \mathbf{u}_j^* + \mathbf{v}, \quad (9)$$

where $\mathbf{v} \sim \mathcal{N}(\mathbf{0}, \sigma^2 \mathbf{I}_{p(N+1)})$, \mathbf{u}_j^* is in (6) and \mathbf{J}_j is in (7).

B. Initial State and Unknown Input Estimation under Sparsity Constraints: A Group LASSO for Approach

Based the measurement model in (9), we introduce the group LASSO estimator to estimate $(\mathbf{x}_0^*, \mathbf{u}_1^*, \dots, \mathbf{u}_m^*)$ and also the active set S . Let $\mathbf{J} = [\mathbf{J}_1^T, \dots, \mathbf{J}_m^T]$ and $\mathbf{u} = [\mathbf{u}_1^T, \dots, \mathbf{u}_m^T]$, where $\mathbf{u}_j \in \mathbb{R}^{N+1}$. Recall the definition of $\ell_{p,0}$ -norm from the notation section, and consider

$$\begin{bmatrix} \hat{\mathbf{x}}_0 \\ \hat{\mathbf{u}} \end{bmatrix} = \arg \min_{\mathbf{x}_0, \mathbf{u}} \left\{ \frac{1}{2T} \|\mathbf{y} - \mathbf{O}\mathbf{x}_0 - \mathbf{J}\mathbf{u}\|_2^2 + \lambda_T \|\mathbf{u}\|_{p,0} \right\}, \quad (10)$$

where the regularization parameter $\lambda_T \geq 0$ and $T = p(N+1)$ is the dimension of \mathbf{y} in (9). The above problem is called subset (or block-column) selection problem because the optimization problem amounts to finding \mathbf{J}_j that contributes to \mathbf{y} in (9).

Unfortunately, (10) is a combinatorial optimization problem and its computationally complexity is exponential in m . We circumvent this difficulty by replacing the $\|\mathbf{u}\|_{p,0}$ with the $\|\mathbf{u}\|_{p,1}$ -norm. This is a common relaxation technique widely used in the literature of compressed sensing and statistics; see [24], [25]. Thus, we end up with the group LASSO problem:

$$\begin{bmatrix} \hat{\mathbf{x}}_0 \\ \hat{\mathbf{u}} \end{bmatrix} \in \arg \min_{\mathbf{x}_0, \mathbf{u}} \left\{ \frac{1}{2T} \|\mathbf{y} - \mathbf{O}\mathbf{x}_0 - \mathbf{J}\mathbf{u}\|_2^2 + \lambda_T \|\mathbf{u}\|_{p,1} \right\}. \quad (11)$$

For definiteness, we set $p = 2$, although our analysis extends to the case $p \neq 2$. In the literature, $\|\mathbf{u}\|_{2,1} = \sum_{j=1}^m \|\mathbf{u}_j\|_2$ is referred to as the block or group norm. Our optimization problem in (11) differs from the traditional group LASSO [6] because we do not penalize \mathbf{x}_0 . This is subtle yet important distinction because in many applications, including power systems, initial state is rarely sparse. In Section VI, we provide details on how to numerically solve (11). Instead, in Section III, for a specific range of λ_T , we show that the group-norm based regularizer promotes group sparsity in $\hat{\mathbf{u}}$ and that $\hat{S} = S$ holds with high probability, where $\hat{S} \triangleq \{j : \hat{\mathbf{u}}_j \neq 0\}$.

Due to the presence of additive noise in the measurement vector \mathbf{y} in (9), neither the estimate $\hat{\boldsymbol{\beta}} = (\hat{\mathbf{x}}_0, \hat{\mathbf{u}})$ in (11) need to identically match $\boldsymbol{\beta}^* = (\mathbf{x}_0^*, \mathbf{u}^*)$ nor does $\hat{S} = S$. Thus, we evaluate the quality of our estimates (i.e., the hatted quantities) in a probabilistic sense using the error metrics:

- $\hat{\boldsymbol{\beta}}$ is said to be ℓ_2 -consistent if $\|\hat{\boldsymbol{\beta}} - \boldsymbol{\beta}^*\|_2 \leq o(T)$ with probability at least $1 - c_1 \exp(-c_2 T)$, for some $c_1, c_2 > 0$.
- $\hat{\mathbf{u}}$ is said to be location recovery consistent if $\hat{S} = S$ with probability at least $1 - c_3 \exp(-c_4 T)$, for $c_3, c_4 > 0$.

Here $o(T)$ implies that the upper bound on the error tends to zero as $T \rightarrow \infty$. The ℓ_2 -error bound ensures that the estimate $\hat{\boldsymbol{\beta}} \approx \boldsymbol{\beta}^*$ by increasing $T = p(N+1)$. Instead, the location selection consistency ensures that as long as T is sufficiently large, \hat{S} correctly identifies the true sources of FOs.

III. DELAYED ESTIMATION AND INVARIANT ZEROS

In this section we cull recent results on the initial state and delayed input recovery using finite number of measurements [26], by assuming the knowledge set S . These results provide a starting point to prove our main results in Section IV

We begin by expressing \mathbf{y} in (9) in a slightly different way. From (5), we have $\mathbf{B}\mathbf{u}^*[k] = \mathbf{B}_S \mathbf{u}_S^*[k] + \sum_{j \in S^c} \mathbf{b}_j u_j^*[k]$.

Substituting this fact in (3) and recursively expanding $\mathbf{y}[k]$ in (4) yields us the following model for \mathbf{y} defined in (6).

$$\mathbf{y} = \underbrace{[\mathbf{O} \quad \mathbf{J}_S]}_{\triangleq \Psi_S} \underbrace{\begin{bmatrix} \mathbf{x}_0^* \\ \mathbf{u}_S^* \end{bmatrix}}_{\triangleq \beta_S^*} + \sum_{j \in S^c} \mathbf{J}_j \mathbf{u}_j^* + \mathbf{v}, \quad (12)$$

where \mathbf{u}_S^* and $\mathbf{J}_S \in \mathbb{R}^{p(N+1) \times m^*(N+1)}$ are defined as

$$\mathbf{u}_S^* = \begin{bmatrix} \mathbf{u}_S^*[0] \\ \mathbf{u}_S^*[1] \\ \mathbf{u}_S^*[2] \\ \vdots \\ \mathbf{u}_S^*[N] \end{bmatrix}; \mathbf{J}_S = \begin{bmatrix} \mathbf{H}_0^{(S)} & \mathbf{0} & \mathbf{0} & \dots & \mathbf{0} \\ \mathbf{H}_1^{(S)} & \mathbf{H}_0^{(S)} & \mathbf{0} & \dots & \mathbf{0} \\ \mathbf{H}_2^{(S)} & \mathbf{H}_1^{(S)} & \mathbf{H}_0^{(S)} & \dots & \mathbf{0} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ \mathbf{H}_N^{(S)} & \mathbf{H}_{N-1}^{(S)} & \dots & \mathbf{H}_1^{(S)} & \mathbf{H}_0^{(S)} \end{bmatrix}, \quad (13)$$

with $\mathbf{H}_0^{(S)} = \mathbf{0}_{p \times m^*}$ and $\mathbf{H}_l^{(S)} = \mathbf{C}\mathbf{A}^{l-1}\mathbf{B}_S$, for all $l \geq 1$. Note that \mathbf{y} in (9) and (12) are exactly the same. Importantly, \mathbf{u}_S^* in (13) is a concatenation of inputs $\mathbf{u}_S^*[k]$ associated with S from $k = 0$ (top) to N (bottom), but not a concatenation of \mathbf{u}_j^* in (6), for all $j \in S$.

To show that the group LASSO is location recovery consistent, or $\hat{S} = S$ holds with high probability, $\Psi_S = [\mathbf{O} \quad \mathbf{J}_S]$ in (12) should be of full column rank. To see this, suppose that $\sigma^2 \approx 0$ and that we know S . Then, by substituting $\mathbf{u}_j^* = \mathbf{0}$, for all $j \in S^c$, and $\mathbf{v} = \mathbf{0}$ in \mathbf{y} in (12), it follows that

$$\mathbf{y} = \Psi_S \beta_S^*. \quad (14)$$

Thus for a rank deficient Ψ_S , we cannot perfectly recover $\beta_S^* = (\mathbf{x}_0^*, \mathbf{u}_S^*[0], \dots, \mathbf{u}_S^*[N])$ even with noise-free measurements and with the knowledge of S . However, unfortunately, unlike the model matrices, such as random design and Fourier basis matrices, considered in signal processing and statistics applications, Ψ_S could be rank deficient. This is so because system in (3)-(4) may not be initial state and input observable [8]; that is, either \mathbf{O} or \mathbf{J}_S is rank deficient, or both \mathbf{O} and \mathbf{J}_S have full ranks, but $[\mathbf{O} \quad \mathbf{J}_S]$ is rank deficient.

From the foregoing discussion, it is clear that recovering β_S^* and full rank of Ψ_S are intimately connected. Interestingly, for d -delay invertible linear systems, even when β_S^* is not recoverable, a portion of it is perfectly recoverable [8], [26]. In fact, we can recover $\beta_{S,[0:N-d]}^* = (\mathbf{x}_0^*, \mathbf{u}_S^*[0], \dots, \mathbf{u}_S^*[N-d])$, where $N \geq d$, from $\mathbf{y}^T = [\mathbf{y}^T[0], \dots, \mathbf{y}^T[N]]$. Here, $d \geq 0$ is called *delay* and we refer $\beta_{S,[0:N-d]}^*$ to as the delayed input. As a result, we show that a specific sub-matrix of Ψ_S has full column rank even when Ψ_S is rank deficient.

We formalize the notion of d -delay. Let $\mathbf{x}_0 = \mathbf{0}$ to note that $\Psi_S = \mathbf{J}_S$ and $\beta_S^* = \mathbf{u}_S^*$. Substituting \mathbf{J}_S (13) in (14), yields

$$\underbrace{\begin{bmatrix} \mathbf{y}[0] \\ \mathbf{y}[1] \\ \vdots \\ \mathbf{y}[N] \end{bmatrix}}_{\mathbf{y}_N} = \underbrace{\begin{bmatrix} \mathbf{H}_0^{(S)} & \mathbf{0} & \dots & \mathbf{0} \\ \mathbf{H}_1^{(S)} & \mathbf{H}_0^{(S)} & \dots & \mathbf{0} \\ \vdots & \vdots & \ddots & \vdots \\ \mathbf{H}_N^{(S)} & \mathbf{H}_{N-1}^{(S)} & \dots & \mathbf{H}_0^{(S)} \end{bmatrix}}_{\triangleq \mathbf{J}_{S,[N:0]}} \underbrace{\begin{bmatrix} \mathbf{u}_S^*[0] \\ \mathbf{u}_S^*[1] \\ \vdots \\ \mathbf{u}_S^*[N] \end{bmatrix}}_{\mathbf{u}_{S,[0:N]}^*}. \quad (15)$$

Notice that $\mathbf{J}_S = \mathbf{J}_{S,[N:0]}$ and $\mathbf{u}_S^* = \mathbf{u}_{S,[0:N]}^*$. Define

$$\mathbf{J}_{S,[N:0]} = \left[\mathbf{M}_N^{(S)} \mid \mathbf{M}_{N-1}^{(S)} \mid \dots \mid \mathbf{M}_0^{(S)} \right], \quad (16)$$

where $\mathbf{M}_l^{(S)}$ denotes the l^{th} block column of $\mathbf{J}_{S,[N:0]}$ labeled right ($l = 0$) to left ($l = N$). By construction $\mathbf{J}_{S,[N:0]}$ is rank deficient because $\mathbf{H}_0^{(S)} = \mathbf{0}$ in (15). Thus, we cannot recover $\mathbf{u}_S^*[N]$ using \mathbf{y}_N . Further, in several practical applications, $\mathbf{H}_1^{(S)} = \mathbf{C}\mathbf{B}_S = \mathbf{0}$ (or has non-full column rank). This is because sensors may not be located at the inputs. For e.g., in power systems, bus level PMUs do not directly measure PSS's output. Thus it is impossible to recover $\mathbf{u}_S^*[N-1]$ using \mathbf{y}_N .

Definition 1. (System delay) For a non-negative integer $d \geq 0$, let $\mathbf{J}_{S,[d:0]}$ be defined as in (15). System in (3)-(4), with $\mathbf{x}_0 = \mathbf{0}$, $\mathbf{u}_{S^c}^* = \mathbf{0}$ and $\sigma^2 = 0$, is d -delay left invertible if

$$\text{Rank}(\mathbf{J}_{S,[d:0]}) - \text{Rank}(\mathbf{J}_{S,[d-1:0]}) = m^*, \quad (17)$$

for $\mathbf{J}_{S,[d:0]}$ defined in (15) and m^* is the dimension of $\mathbf{u}_S^*[k]$. The smallest d that satisfies (17) is denoted as η_S . \square

Throughout we assume $d = \eta_S$ and set $d \triangleq \infty$ if (17) does not hold for any $d \geq 0$. Suppose that $\eta_S < \infty$. Then, from the rank properties of partitioned matrices [26], it follows that $\mathbf{M}_d^{(S)}$ in (15) is full column rank. Thus, there exists a matrix \mathbf{S} such that $\mathbf{S}\mathbf{y}_d = \mathbf{u}_S^*[0]$. We may recover $\mathbf{u}_S^*[1]$ using the residual $\hat{\mathbf{y}}_{d+1} \triangleq \mathbf{y}_{d+1} - \mathbf{M}_{d+1}^{(S)}\mathbf{u}_S^*[0]$. In fact, $\mathbf{S}\hat{\mathbf{y}}_{d+1} = \mathbf{u}_S^*[1]$. By iterating this procedure, we can recover inputs in $\mathbf{u}_{S,[0:N-d]}^* \triangleq [(\mathbf{u}_S^*[0])^T, \dots, (\mathbf{u}_S^*[N-d])^T]^T$ using \mathbf{y}_N .

We relax $\mathbf{x}_0 = \mathbf{0}$ assumption and extend the rank condition in (17) to recover jointly $\beta_{S,[0:N-d]}^* = (\mathbf{x}_0^*, \mathbf{u}_{S,[0:N-d]}^*)$, as a whole rather than sequentially, using \mathbf{y}_N . First, we define the smallest delay for recovering \mathbf{x}_0^* in the presence of input:

$$\mu_S \triangleq \min\{d \geq 0 : \text{Rank}([\mathbf{O}_d \quad \mathbf{J}_{S,[d:0]}) - \text{Rank}(\mathbf{J}_{S,[d:0]}) = n\}, \quad (18)$$

where $\mathbf{O}_d = [\mathbf{C}^T (\mathbf{C}\mathbf{A})^T \dots, (\mathbf{C}\mathbf{A}^d)^T]^T$, and n is the dimension of \mathbf{A} . The rank condition in (18) says that \mathbf{O}_d has full column rank ($= n$) and that the columns in \mathbf{O}_d are linearly independent of columns in $\mathbf{J}_{S,[d:0]}$. This condition is stronger than system in (3)-(4) being observable, as shown below:

Example 1. Let $\mathbf{A} = [1 \quad 2; 0 \quad 3]$, $\mathbf{B}_S = [2 \quad 3]^T$, and $\mathbf{C} = [1 \quad 0]$. Then $\eta_S = 1$ and $\text{Rank} \mathbf{O}_l = 2$, for any $l \geq 2$; that is, the system is observable. However, $\mu_S = \infty$ because the rank condition in (18) does not hold. This is to be expected because the second column of \mathbf{A} is identical to \mathbf{B}_S . \square

Let $\mathbf{M}_l^{(S)}$ be as in (16). For $N \geq d \geq 0$, consider

$$\underbrace{[\mathbf{O} \quad \mathbf{J}_S]}_{\Psi_S} = \underbrace{[\mathbf{O} \quad \mathbf{M}_N^{(S)} \quad \dots \quad \mathbf{M}_d^{(S)}]}_{\Psi_{S,[N:d]}} \underbrace{[\mathbf{M}_{d-1}^{(S)} \quad \dots \quad \mathbf{M}_0^{(S)}]}_{\Psi_{S,[d-1:0]}}. \quad (19)$$

Let Ψ_S^+ be the pseudo inverse of Ψ_S . The proposition below states conditions under which we can recover $(\mathbf{x}_0^*, \mathbf{u}_{S,[0:N-d]}^*)$.

Proposition 1. Suppose that η_S in (17) and μ_S in (18) are finite. Then, for $N \geq \max\{\eta_S, \mu_S\}$ with $d \geq \eta_S$, we have

- 1) $\Psi_{S,[N:d]}$ defined in (19) has full column rank.
- 2) $\mathcal{R}(\Psi_{S,[N:d]}) \cap \mathcal{R}(\Psi_{S,[d-1:0]}) = \{\mathbf{0}\}$.

Moreover, for $t_S \triangleq (N - d + 1)m^*$ and $m^* = |S|$, we have

$$\begin{bmatrix} \mathbf{x}_0^* \\ \tilde{\mathbf{u}}_{S,[0:N-d]}^* \end{bmatrix} = \underbrace{\begin{bmatrix} \mathbf{I}_{n+t_S} & \mathbf{0}_{(n+t_S) \times dm^*} \end{bmatrix}}_{\tilde{\mathbf{\Pi}}_{S,[0:N-d]}} \Psi_S^+ \mathbf{y}. \quad (20)$$

The proof of this fact is given in [26, Theorem 7]. Part (1) of proposition states that the sub-matrix $\Psi_{S,[N:d]}$ has full rank even when $\Psi_{S,[N:0]}$ is rank deficient. This fact plays a vital role in the performance analysis of the group LASSO estimate.

For Proposition 1 to hold, we require $\eta_S, \mu_S < \infty$. Using the notion of zeros and rank of the system matrix (see below), we state verifiable conditions to check if $\eta_S, \mu_S < \infty$. For all $z \in \mathbb{C}$, define the system and transfer matrix:

$$\mathcal{Z}_S[z] \triangleq \begin{bmatrix} z\mathbf{I} - \mathbf{A} & -\mathbf{B}_S \\ \mathbf{C} & \mathbf{0} \end{bmatrix} \text{ and} \quad (21)$$

$$\mathcal{G}_S[z] \triangleq \mathbf{C}(z\mathbf{I} - \mathbf{A})^{-1}\mathbf{B}_S, \quad z \notin \text{spec}(\mathbf{A}), \quad (22)$$

where $\text{spec}(\mathbf{A})$ is the multiset of eigenvalues of \mathbf{A} . Define the normal ranks of $\mathcal{Z}_S[z]$ and $\mathcal{G}_S[z]$, respectively, as $\text{nRank} \mathcal{Z}_S \triangleq \max_{z \in \mathbb{C}} \text{Rank} \mathcal{Z}_S[z]$ and $\text{nRank} \mathcal{G}_S \triangleq \max_{z \in \mathbb{C}} \text{Rank} \mathcal{G}_S[z]$. A number $z_0 \in \mathbb{C}$ is called the invariant zero of $(\mathbf{A}, \mathbf{B}_S, \mathbf{C})$ if $\text{Rank} \mathcal{Z}_S[z_0] < \text{nRank} \mathcal{Z}_S$. If $(\mathbf{A}, \mathbf{B}_S, \mathbf{C})$ has invariant zeros, there exists $\mathbf{u}_S^* \neq \mathbf{0}$ and $\mathbf{x}_0 \neq \mathbf{0}$ such that (noise-free) $y[k] = \mathbf{0}$, for all $k \geq 0$ [27]. (Thus, we cannot distinguish between non-zero and zero inputs from \mathbf{y}_N alone.) Hence, $\eta_S, \mu_S = \infty$.

Lemma 2. *Let $(\mathbf{A}, \mathbf{B}_S, \mathbf{C})$ has no invariant zeros. Then, (i) $\mu_S < \infty$ and, for $N \geq \mu_S$, system in (3)-(4) is initial state observable; and (ii) if $\text{nRank} \mathcal{Z}_S = n + m^*$, $\eta_S < \infty$.*

A proof for the statement (i) can be found in [26, Proposition 5]. Instead, the statement (ii) follows from [8, Theorem 1, pp. 227]. Thus, if $(\mathbf{A}, \mathbf{B}_S, \mathbf{C})$ satisfies conditions in Lemma 2 the assumptions in Proposition 1 hold. Hence, the sub-matrix $\Psi_{S,[N:d]}$ has full rank and we can recover $(\mathbf{x}_0^*, \mathbf{u}_{S,[0:N-d]}^*)$.

IV. LOCATION RECOVERY AND ESTIMATION CONSISTENCY OF THE GROUP LASSO ESTIMATOR

We theoretically investigate the performance of the group LASSO estimator in (11) using the previously stated results for the delayed input estimation. Our results generalize the existing group LASSO's guarantees for static (or non-dynamical) systems [5], [28] to the dynamical systems with delay $d \geq 0$.

Recall that the estimate in (11) is $(\hat{\mathbf{x}}_0, \hat{\mathbf{u}}_1, \dots, \hat{\mathbf{u}}_m)$, where $\hat{\mathbf{u}}_j = [\hat{u}_j[0], \dots, \hat{u}_j[N]]^T$. For any $R \subset \{1, \dots, m\}$, we define $\hat{\mathbf{u}}_R[k] = [\hat{u}_{r_1}[k], \dots, \hat{u}_{r_{|R|}}[k]]$, for all $k \geq 0$ and $r_j \in R$. In words, we group the estimated inputs associated with the set R . Further, define $\hat{\mathbf{u}}_R^T = [\hat{\mathbf{u}}_R^T[0], \dots, \hat{\mathbf{u}}_R^T[N]]$. Thus, we can compare $\hat{\beta}_S = (\hat{\mathbf{x}}_0, \hat{\mathbf{u}}_S)$ and $\beta_S^* = (\mathbf{x}_0^*, \mathbf{u}_S^*)$ in (12).

Recall that $S = \{j : \mathbf{u}_j^* \neq \mathbf{0}\}$ and $\hat{S} = \{j : \hat{\mathbf{u}}_j \neq \mathbf{0}\}$. We derive conditions under which (i) $\hat{S} = S$ and (ii) $\|\beta_{S^*}^* - \hat{\beta}_{S^*}\|_2 \leq \epsilon$, for any $\epsilon > 0$, hold with high probability. To this aim, we make a few assumptions. If $d = 0$ and $\mathbf{x}_0^* = \mathbf{0}$, these assumptions reduce to the standard group LASSO's assumptions [25].

Assumption 3. (Identifiability and mutual incoherence conditions) *Consider the following conditions:*

(A1) *Group normalization: The column block matrices \mathbf{O} and \mathbf{J}_i in (7) satisfy the group normalization condition:*

$$\max \{\|\mathbf{O}\|_2, \|\mathbf{J}_1\|_2, \dots, \|\mathbf{J}_m\|_2\} \leq C\sqrt{T} < \infty. \quad (23)$$

(A2) *Least singular value: Let $\Psi_{S,[N:d]}$ and $\Psi_{S,[d-1:0]}$ be as in (19), and define $\mathbf{M} = [\mathbf{I} - \Psi_{S,[d-1:0]} \Psi_{S,[d-1:0]}^+]$. Then*

$$\left\| \left(\frac{\Psi_{S,[N:d]}^T \mathbf{M} \Psi_{S,[N:d]}}{T} \right)^\dagger \right\|_2 \leq \frac{1}{c_{\min}} < \infty. \quad (24)$$

(A3) *Mutual incoherence: There exists some $\alpha \in [0, 1)$, referred to as "mutual incoherence" parameter, such that*

$$\text{MIC} \triangleq \max_{j \in S^c} \left\| \mathbf{J}_j^T \Psi_S (\Psi_S^T \Psi_S)^+ \right\|_2 \leq \alpha/m^*. \quad (25)$$

Assumption (A1) holds for asymptotically stable systems; that is, the eigenvalues of \mathbf{A} completely lie inside a complex unit circle. Assumption (A2) enforces conditions on the model identifiability; that is, the uniqueness of the delayed estimate $\hat{\mathbf{u}}_{S,[0:N-d]}$ but not necessarily on $\hat{\mathbf{u}}_{S,[0:N]}$ as we shall see later. Assumption (A2) is satisfied for d -delay systems because $\Psi_{S,[N:d]}$ has full column rank if $N \geq d$ (see Section III).

Assumption (A3) is satisfied if Ψ_S and \mathbf{J}_j are orthogonal ($\mathbf{J}_j^T \Psi_S = \mathbf{0}$, for all $j \in S^c$). Orthogonality is restrictive as number of inputs can be more than outputs, or any column of \mathbf{B}_S in (3) can be a linear combination of \mathbf{b}_j , for $j \in S^c$. Nonetheless, (A6) imposes a type of "approximate" orthogonality between \mathbf{J}_j , where $j \in S^c$, and Ψ_S . We quantify this approximation using the parameter α . The ℓ_2 -norm bound in (25) could be conservative as the bound depends on m^* . This dependence can be avoided by working with the ℓ_1 -norm bound; that is, $\max_{j \in S^c} \left\| \mathbf{J}_j^T \Psi_S (\Psi_S^T \Psi_S)^+ \right\|_1 \leq \alpha$. However, we stick with (25) as it is useful to derive an upper bound on MIC in (25) using the system transfer function. In simulations, we study the conservatism incurred due to ℓ_2 -norm based MIC.

Theorem 4. (Location recovery consistency) *Consider the model (12) satisfying assumptions (A1)-(A3) with the active set $S = \{1, \dots, m^*\}$. For some $\delta > 0$ suppose that we select*

$$\lambda_T = \frac{\sqrt{32}C\sigma}{1 - \alpha} \left\{ \sqrt{\frac{(N+1)c_1 + \log(m - m_0)}{T}} + \frac{\delta}{2} \right\}, \quad (26)$$

where $c_1 = \log(5)$. Then, the following hold with probability at least $1 - 4\exp(-T\delta^2/2)$.

- (Non-uniqueness): For $d > 0$, there are infinitely many solutions of (11); however, if $d = 0$, $\hat{\beta}$ in (11) is unique.
- (No false inclusion): For all $d \geq 0$, the support set of any optimal estimate $\hat{\beta}$ is contained within the true support set; that is, $\hat{S} \subset S$.
- (ℓ_∞ bounds): The delayed inputs satisfy the following bound: $\max_{j \in S} \|\hat{\mathbf{u}}_{j,[0:N-d]} - \mathbf{u}_{j,[0:N-d]}^*\|_\infty \leq \beta_{\min}$, where

$$\beta_{\min} = \frac{\sigma}{\sqrt{c_{\min}}} \left\{ \sqrt{\frac{2 \log((N-d+1)m^*)}{T}} + \delta \right\} + \lambda_T \left\| \mathbf{\Pi}_{S,[0:N-d]} \left(\frac{\Psi_S^T \Psi_S}{T} \right)^\dagger \right\|_\infty, \quad (27)$$

$\mathbf{\Pi}_{S,[0:N-d]} = [\mathbf{0}_{t_S \times n} \mathbf{I}_{t_S \times dm^*} \mathbf{0}_{t_{Sc} \times t_{Sc}}]$ and $t_S = (N - d + 1)m^*$.

(d) (Minimum input magnitude and no false exclusion): If $\min_{j \in S} \|\mathbf{u}_j^*_{[0:N-d]}\|_\infty \geq \beta_{\min}$, we have $\hat{S} = S$.

Proof. See Appendix. \square

Corollary 5. Consider $\hat{\beta}_{S,[0:N-d]} = (\hat{\mathbf{x}}_0, \hat{\mathbf{u}}_{S,[0:N-d]})$ and $\beta_{S,[0:N-d]}^* = (\mathbf{x}_0^*, \mathbf{u}_{S,[0:N-d]}^*)$. Let $c_1 = \log(5)$ and $t_S = (N - d + 1)m^*$. Under the assumptions of Theorem 4 with probability at least $1 - \exp(-\delta^2 T/2)$, we have

$$\begin{aligned} \left\| \beta_{S,[0:N-d]}^* - \hat{\beta}_{S,[0:N-d]} \right\|_2 \leq \\ \frac{2\sigma}{\sqrt{c_{\min}}} \left\{ \sqrt{\frac{2c_1(n+t_S)}{T}} + \delta \right\} + \lambda_T \sqrt{\frac{m^*}{Tc_{\min}}}, \end{aligned} \quad (28)$$

We use the *primal-dual witness* technique of Wainwright [25], [28], [29] to prove Theorem 4

Part (a) in Theorem 4 states that the group LASSO estimate β is non-unique unless the sub-system realized by $(\mathbf{A}, \mathbf{B}_S, \mathbf{C})$ has zero delay. This is because, for $N > d > 0$, the sub-matrix $\Psi_{S,[N:d]}$ in (19) has full rank, but not Ψ_S . However, Part (b) in Theorem 4 states that $\hat{S} \subseteq S$, for any optimal estimate β in (11). Thus, the estimated inputs restricted to the complement set are zero: $\hat{\mathbf{u}}_{j^c} = \mathbf{0}$, for all $j \in S^c$. Thus, the non-uniqueness of the optimal solution does not effect the location consistency of the group LASSO estimator.

Part (d) in Theorem 4 (d)—a consequence of the ℓ_∞ norm bound in part (b)—says that for $\hat{S} = S$ to hold (i.e., to detect true inputs correctly), the true non-zero input signal strength should not be too small, precisely, smaller than β_{\min} in (27). The probabilistic result in Theorem 4 also helps determine the number of measurements (N) or sensors (p) required to achieve certain amount of performance. Let us simplify λ_T in (26) to comment on its scaling. By substituting $T = p(N+1)$ and assuming that $\log(m - m^*)/(N+1) \gg c_1$, we have

$$\lambda_T = O \left(\sqrt{\frac{\log(m - m^*)}{p(N+1)}} + \frac{\delta}{2} \right). \quad (29)$$

For $p = 1$, λ_T in (29) reduces to that of λ_T for the traditional LASSO problem [25]. Thus, the term $c_1(N+1)/T$ in (26) takes into consideration the number of unknowns in \mathbf{u}_j^* , and p in $p(N+1)$ accounts for the number of sensors.

The choice of λ_T plays an important role in determining if Theorem 4 (c) (that is, $\hat{S} = S$) holds. In fact, the smaller the λ_T , the smaller the minimum threshold β_{\min} . Interestingly, for $\lambda_T = 0$, which happens, say, when $\sigma = 0$, the optimization problem in (11) reduces to the standard ordinary least squares (OLS) problem. Thus, there is no shrinkage of input estimates toward zero. Further, λ_T does not depend on c_{\min} in (24) but depends on the group normalization constant C in (23) and the mutual incoherence parameter α in (25).

To understand the role of C on λ_T , and ultimately on β_{\min} , let $d = 0$ and note that Ψ_S full rank. Assuming (A1) holds with equality, from the standard norm inequalities, we have

$$\kappa_1 + \lambda_T \sqrt{\kappa_2}/C^2 \geq \beta_{\min} \geq \kappa_1 + \lambda_T/(C^2 \sqrt{\kappa_2}),$$

where κ_1 is the first term on the right side of the equality in (27) and $\kappa_2 = (N+1)m^*$ is the dimension of \mathbf{u}_S^* . Noting that λ_T is proportional to C , we see that $\beta_{\min} = \kappa_1 + O(\sqrt{\kappa_2}/C)$. As expected, larger values of C results in smaller β_{\min} because the effective signal strength of $\Psi_S \mathbf{u}_S^*$ is large. Instead, smaller values of C results in higher β_{\min} , thereby requiring \mathbf{u}_S^* to be large. If not, the strength of $\Psi_S \mathbf{u}_S^*$ decreases. Finally, from (26), we observe that λ_T is an increasing function of $\alpha \in [0, 1]$; thus, higher the α larger is the β_{\min} . Recall that α is large if \mathbf{J}_j , for $j \in S^c$, is highly correlated with Ψ_S .

We now comment on the ℓ_2 -error bound between $\beta_{S,[0:N-d]}^*$ and $\hat{\beta}_{S,[0:N-d]}$ given in Corollary 4. First, the error bound depends on the number of unknown parameters $n + t_S = n + (N - d + 1)m^*$, i.e., the dimension of the initial state and delayed input. Letting $T = p(N+1) \gg n$, we observe that the first term of the bound in (28) scales as $O(\tilde{c}(\sqrt{m^*/p} + \delta))$, where $\tilde{c} = 2\sigma/\sqrt{c_{\min}}$. Thus, more PMUs result in less error. However, the bound is loose for large values of λ_T . To remedy this shortcoming, we consider the following OLS estimate:

$$\hat{\beta}_{\hat{S},[0:N-d]}^{(OLS)} \triangleq \tilde{\mathbf{\Pi}}_{\hat{S},[0:N-d]} (\Psi_{\hat{S}}^+ \mathbf{y}), \quad (30)$$

where $\tilde{\mathbf{\Pi}}_{\hat{S},[0:N-d]}$ is defined similar to $\tilde{\mathbf{\Pi}}_{S,[0:N-d]}$ in (20). We present the second main result of this section: an oracle bound on the error $\|\beta_{S,[0:N-d]}^* - \hat{\beta}_{\hat{S},[0:N-d]}^{(OLS)}\|_2$.

Theorem 6. (ℓ_2 -consistency: oracle bounds) Suppose that the hypotheses in Theorem 4 hold. Then, for any $\delta, \delta_1 > 0$, with probability at least $1 - 4 \exp(-T\delta^2/2) - \delta_1$,

$$\begin{aligned} \left\| \beta_{S,[0:N-d]}^* - \hat{\beta}_{\hat{S},[0:N-d]}^{(OLS)} \right\|_2 \leq \frac{4\sigma}{\sqrt{c_{\min}}} \left\{ \sqrt{\frac{(n+t_S)}{T}} \right\} \\ + \frac{2\sigma}{\sqrt{c_{\min}}} \left\{ \sqrt{\frac{1}{T} \log\left(\frac{1}{\delta_1}\right)} \right\}, \end{aligned} \quad (31)$$

The proof is in Appendix. Similar to the bound in Corollary 5, the first term in (31) is $O(\tilde{c}(\sqrt{m^*/p}))$; however, the second term in (31) does not depend on λ_T and it approaches zero as $T \rightarrow \infty$. Thus, the overall error is dictated by m^*/p . We call the bound in (31) as the oracle because the bound holds for $\hat{\beta}_{S,[0:N-d]}^{(OLS)}$, albeit with probability $1 - \delta_1$.

A. Extensions of group LASSO guarantees to noisy dynamics

We extend our results in Theorems 4 and 6 to the setting where system in (3)-(4) is affected by both state and measurement noises. We also relax the diagonal covariance structure of the measurement noise. Consider the following dynamics:

$$\begin{aligned} \mathbf{x}[k+1] &= \mathbf{A}\mathbf{x}[k] + \mathbf{B}_S \mathbf{u}_S^*[k] + \sum_{j \in S^c} \mathbf{b}_j u_j^*[k] + \mathbf{w}[k] \\ \mathbf{y}[k] &= \mathbf{C}\mathbf{x}[k] + \mathbf{v}[k], \end{aligned} \quad (32)$$

where the noise random vectors $\mathbf{w}[k] \stackrel{iid}{\sim} \mathcal{N}(\mathbf{0}, \mathbf{Q})$ and $\mathbf{v}[k] \stackrel{iid}{\sim} \mathcal{N}(\mathbf{0}, \mathbf{R})$, with $\mathbf{Q} \succeq \mathbf{0}$ and $\mathbf{R} \succ \mathbf{0}$, are uncorrelated. Let $\mathbf{y} = [\mathbf{y}[0]^T \dots \mathbf{y}[N]^T]^T$, and from (32), note that

$$\mathbf{y} = \mathbf{O}\mathbf{x}_0^* + \mathbf{J}_S \mathbf{u}_S^* + \sum_{j \in S^c} \mathbf{J}_j \mathbf{u}_j^* + \mathbf{J}_w \mathbf{w} + \mathbf{v}, \quad (33)$$

where $\mathbf{w} \triangleq [\mathbf{w}[0]^\top \dots \mathbf{w}[N]^\top]^\top$ and $\mathbf{v} = [\mathbf{v}[0]^\top \dots \mathbf{v}[N]^\top]^\top$. The noise response matrix \mathbf{J}_w is obtained by replacing $\mathbf{H}_k^{(1)}$ in \mathbf{J}_1 , given by (7), with $\mathbf{C}\mathbf{A}^{k-1}$, for all $k \geq 0$. Because \mathbf{w} and \mathbf{v} are Gaussian, it follows that $\mathbf{J}_w \mathbf{w} + \mathbf{v} \sim \mathcal{N}(\mathbf{0}, \Sigma_{\tilde{\mathbf{v}}})$, where $\Sigma_{\tilde{\mathbf{v}}} = [\mathbf{J}_w \mathbf{w} + \mathbf{v}][\mathbf{J}_w \mathbf{w} + \mathbf{v}]^\top$. Finally, define $\tilde{\sigma}^2 = \|\Sigma_{\tilde{\mathbf{v}}}\|_2$.

Suppose that we solve the group LASSO problem in (11) for the model in (33). Then, Theorems in (4) and (6) hold true for $\sigma^2 = \tilde{\sigma}^2$. However, the modified noise variance ($\tilde{\sigma}^2$) could be large depending on the system matrices in (32).

B. Mutual Incoherence: Frequency Domain

Thus far we discussed the location recovery- and estimation-consistency of the group LASSO estimator in (11) assuming that assumptions in (A1)-(A3) hold of which the first two are satisfied by stable dynamical systems with $(\mathbf{A}, \mathbf{B}_S, \mathbf{C})$ having no invariant zeros². However, (A3) might not hold for arbitrary systems, and moreover, verifying (25) can be computationally demanding when either N (the measurement horizon) or n (dimension of system matrix \mathbf{A}) is large. In what follows, we bound $\max_{j \in S^c} \|\mathbf{J}_j^\top \Psi_S (\Psi_S^\top \Psi_S)^+\|_2$ in (25) using a quantity that depends on the transfer function matrices associated with $(\mathbf{A}, \mathbf{B}_S, \mathbf{C})$ and $(\mathbf{A}, \mathbf{b}_j, \mathbf{C})$, for $j \in S^c$. The advantage is that this upper bound can be computed efficiently, as it depends only on the lower dimensional system matrices but not on N .

To simplify the exposition, we assume $\mathbf{x}_0 = \mathbf{0}$; thus, $\Psi_S = \mathbf{J}_S$. Similar to the transfer matrix $\mathcal{G}_S[z]$ in (22), define $\mathcal{G}_j[z] = \mathbf{C}(z\mathbf{I} - \mathbf{A})^{-1} \mathbf{b}_j$ and $\mathcal{G}_{S^c}[z] = \mathbf{C}(z\mathbf{I} - \mathbf{A})^{-1} \mathbf{B}_{S^c}$, where \mathbf{B}_{S^c} is the matrix composed of columns \mathbf{b}_j , with $j \in S^c$.

Theorem 7. *Assumption (A6) holds if $\text{rank} \mathcal{Z}_S = n + m^*$ and*

$$\max_{j \in S^c} \max_{\{z \in \mathbb{C}: |z|=1\}} \|\mathcal{G}_S^+[z] \mathcal{G}_j[z]\|_2 \leq \alpha/m^* < 1. \quad (34)$$

Proof. See Appendix. \square

We refer to the expression in (34) as the frequency domain mutual incoherence condition. Thus to verify Assumption (A6), we need to check if the worst case gain of the transfer matrix $\mathcal{G}_S^+[z] \mathcal{G}_j[z]$ is bounded above by α/m^* ; see Fig. 1. If computing (34) is prohibitive for each $j \in S^c$, we can resort to the weaker condition: $\max_{\{z \in \mathbb{C}: |z|=1\}} \|\mathcal{G}_S^+[z] \mathcal{G}_{S^c}[z]\|_2 \leq \alpha/m^* < 1$. To appreciate the condition in (34), we take \mathcal{Z} -transform of system in (3)-(4), and then note that

$$\mathbf{y}[z] = \mathcal{G}_S[z] \mathbf{u}_S[z] + \sum_{j \in S^c} \mathcal{G}_j[z] \mathbf{u}_j[z], \quad \forall z \notin \text{spec}(\mathbf{A}).$$

By pre-multiplying the above identity with $\mathcal{G}_S^+[z]$, we have

$$\begin{aligned} \mathcal{G}_S^+[z] \mathbf{y}[z] &= \mathbf{u}_S[z] + \sum_{j \in S^c} \mathcal{G}_S^+[z] \mathcal{G}_j[z] \mathbf{u}_j[z] \quad \forall z \notin \text{spec}(\mathbf{A}) \\ &= \mathbf{u}_S[z] + \mathcal{G}_S^+[z] \mathcal{G}_{S^c}[z] \mathbf{u}_{S^c}[z], \end{aligned}$$

Thus to recover $\mathbf{u}_S[z]$ accurately, the gain $\|\mathcal{G}_S^+[z] \mathcal{G}_j[z]\|_2$ or $\|\mathcal{G}_S^+[z] \mathcal{G}_{S^c}[z]\|_2$ needs to be small.

We conclude this section by highlighting three special cases where (34) holds: (i) $\mathcal{R}(\mathcal{G}_{S^c}[z]) \subseteq \mathcal{R}^\perp(\mathcal{G}_S^+[z]) = \mathcal{R}^\perp(\mathcal{G}_S^T[z])$, for all $|z| = 1$. In other words, the columns of $\mathcal{G}_{S^c}[z]$ lie in the

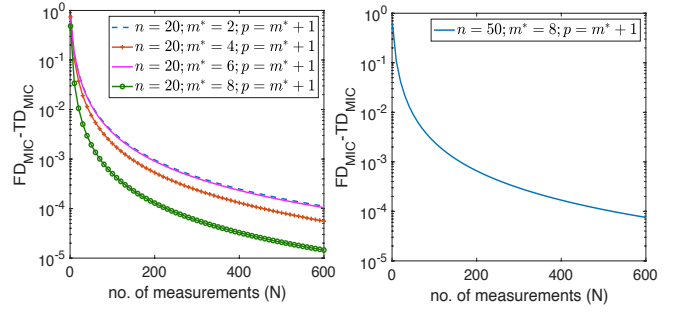


Fig. 1. Illustration of Theorem 7 for system matrices generated using `drss` command in MATLAB. The number of possible sources $m = 10$. In both panels, the y-axis, $FD_{\text{MIC}} - TD_{\text{MIC}}$, is the error between frequency- and time-domain mutual incoherences. (Left panel) We fix $n = 20$ and plot $FD_{\text{MIC}} - TD_{\text{MIC}}$ for several values of m^* . (Right panel) For a large dimensional matrix \mathbf{A} , we fix m^* and p , and plot $FD_{\text{MIC}} - TD_{\text{MIC}}$ for several values of system dimension n . In both the panels, the error is positive and is monotone in N implying that $FD_{\text{MIC}} \geq TD_{\text{MIC}}$, as predicted by Theorem 7

left nullspace of $\mathcal{G}_S[z]$; (ii) $\mathcal{G}[z] = [\mathcal{G}_S[z] \quad \mathcal{G}_{S^c}[z]]$ is all-pass³ and (iii) $\mathcal{G}_{S^c}[z] = \alpha \mathcal{G}_S[z]$. The first two cases are rather strong and does not allow columns of $\mathcal{G}_{S^c}[z]$ to be in the range space of $\mathcal{G}_S[z]$. Instead, (ii) models another extreme where the range spaces of $\mathcal{G}_S[z]$ and $\mathcal{G}_{S^c}[z]$ are aligned with each, modulo the factor $\alpha \in [0, 1)$. The latter case in the compressed sensing literature is referred to as overcomplete dictionaries [30].

V. SIMULATIONS

We illustrate the performance of the group LASSO estimator on a large-scale power network and a random system. The following proposition states that the unknown input and initial state can be estimated in two stages. Consequently, we use off-the-shelf ADMM [9] to estimate the input first and then use this estimate to compute the initial state.

Proposition 8. *Suppose that system in (3)-(4) is observable. The optimization problem (11) is equivalent to*

$$\hat{\mathbf{u}} = \arg \min_{\mathbf{u} \in \mathbb{R}^{mT}} \frac{1}{2T} \|\mathbf{\Pi}(\mathbf{y} - \mathbf{J}\mathbf{u})\|_2^2 + \lambda_T \sum_{j=1}^m \|\mathbf{u}_j\|_2, \quad (35)$$

$$\hat{\mathbf{x}}_0 = \mathbf{O}^+(\mathbf{y} - \mathbf{J}\hat{\mathbf{u}}), \quad (36)$$

where $\mathbf{O}^+ = (\mathbf{O}^\top \mathbf{O})^{-1} \mathbf{O}^\top$ and $\mathbf{\Pi} = \mathbf{I} - \mathbf{O}\mathbf{O}^+$.

The proof follows from the KKT conditions [31]. The inputs to the ADMM [9] are the system matrices $(\mathbf{A}, \mathbf{B}, \mathbf{C})$, the measurement \mathbf{y} , and the tuning parameter $\lambda_T \geq 0$. Finally, we note that the two-stage estimation method is one way to implement the group LASSO numerically. One may also use other numerical algorithms to estimate $(\mathbf{x}_0^*, \mathbf{u}^*)$ in one shot.

We evaluate the group LASSO estimator's localization performance using the false-positive rate (FPR) := $|S^c \cap \hat{S}|/|S^c|$, the false-negative rate (FNR) := $|S \cap \hat{S}^c|/|S|$, and the exact recovery rate (ERR) := $(|S \cap \hat{S}| + |S^c \cap \hat{S}^c|)/m$. Recall that FPR and FNR, respectively, measure the proportion of inputs that are falsely identified and left out. Instead, we quantify the estimation performance using the error metrics:

²Systems having invariant zeros lie in a zero measure set [27].

³A real rational transfer function matrix $G[z]$ is all-pass if $G[z]G[1/z] = \mathbf{I}$.

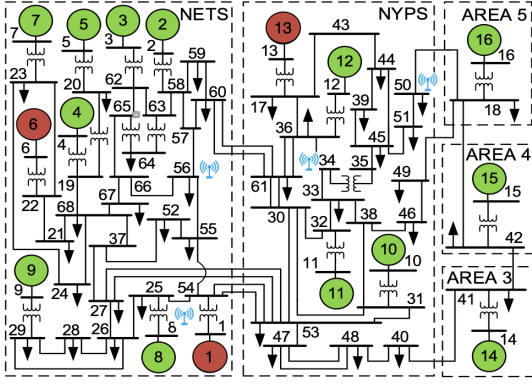


Fig. 2. IEEE 16 machine 68 bus system [33]. Circles, arrows, and curly windings, respectively, denote generator buses, load buses, and transformers. The FO input enters through set points of AVRs associated with the generators at buses $\{1, 6, 13\}$ (red circles). Sensors are located at buses $\{8, 34, 50, 56\}$.

$\|\mathbf{x}_0^* - \hat{\mathbf{x}}_0\|_2 / \|\mathbf{x}_0^*\|_2$ and $\|\mathbf{u}^* - \hat{\mathbf{u}}\|_2 / \|\mathbf{u}^*\|_2$. For the test cases below, the results are averaged over 50 runs.

(Power system) We apply our estimator in (35) to localize the sources of forced oscillatory (FO) inputs in the IEEE 68 bus system 16 machine system (see Fig. 2). Each machine (or generator) consists of ten states, including rotor angle, speed, and the states of the AVR (automatic voltage regulator) and PSS. We model FOs as inputs injected by the AVRs and use bus voltage magnitudes as measurements. For the sampling time $\delta t = 0.1$, we obtained the system matrices $\mathbf{A} \in \mathbb{R}^{160 \times 160}$, $\mathbf{B} \in \mathbb{R}^{160 \times 16}$, and $\mathbf{C} \in \mathbb{R}^{p \times 160}$, where $p \leq 68$, using the Power System Toolbox [32]. Among $m = 16$ possible inputs, we assume $m^* = 3$ with the following inputs: $u_1^*[k] = 0.5 \sin[(2\pi f \delta t)k] + w[k]$, $u_6^*[k] = 0.6 \sin[(2\pi f \delta t)k] + w[k]$, and $u_{13}^*[k] = 0.7 \sin[(2\pi f \delta t)k] + w[k]$, where $f = 1.5 \mathcal{U}(0, 1)$ and $w[k] \sim \mathcal{N}(0, 0.05^2)$. We set $p = 4$ and choose sensor locations arbitrarily with the only exception that these are non-collocated with inputs (shown in Fig. 2). Let $\mathbf{x}_0 = \mathbf{0}$ (the non-zero case is considered in the subsequent case). Finally, we let $N = 100$ and the noise variance $\sigma^2 = 0.01$.

In Fig. 3, we plot the FPR, FNR, and ERR with respect to λ_T . As expected, the FNR increases with λ_T whereas the FPR decreases with λ_T , although not monotonically. From the bottom left panel, we can infer that values of $\lambda_T \in (0.3, 0.4)$ yield maximum ERR. In the bottom right panel, note that for $\lambda_T = 0.288$, the group LASSO estimator accurately localized inputs among 40 out of 50 runs. In Fig. 4, for a measurement realization where the group LASSO estimator identified true locations, we plot the inputs estimated by the group LASSO and the reduced model based OLS estimators.

(Large-scale random system) Following [19], we generate matrices as follows: $\mathbf{A}_{ij} \stackrel{\text{iid}}{\sim} \mathcal{N}(0, 1/n)$; $\mathbf{C}_{ij} \stackrel{\text{iid}}{\sim} \mathcal{N}(0, 1)$; and $\mathbf{B}^T = [\mathbf{I}_m^T \quad \mathbf{0}^T]$. We let $\mathbf{x}_0 \sim \mathcal{N}(\mathbf{0}, \mathbf{I}_n)$ and the measurement noise variance parameter $\sigma = 0.01$. We set $n = 50$, $m = 30$, and $m^* = 5$. The active set $S = \{1, 2, 3, 4, 5\}$ and $u_j[k]$ is sampled uniformly on $[-2, 2]$, for all $j \in S$ and $k \in [N]$. The sensors measures the first $p (\leq n)$ states. In Fig. 5 for $p = 15$, we plot the average estimation error metrics as a function of the measurement horizon (N). In both the panels, estimation errors remain uniform across N because the number of (to be

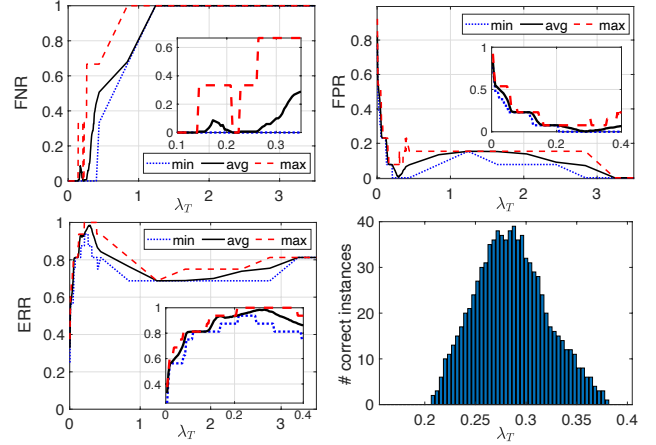


Fig. 3. False negative rate (FNR), false positive rate (FPR), exact recovery rate (ERR), and the number of exactly recovered instances among 50 runs of the IEEE 16 machine 68 bus system data using the group LASSO.

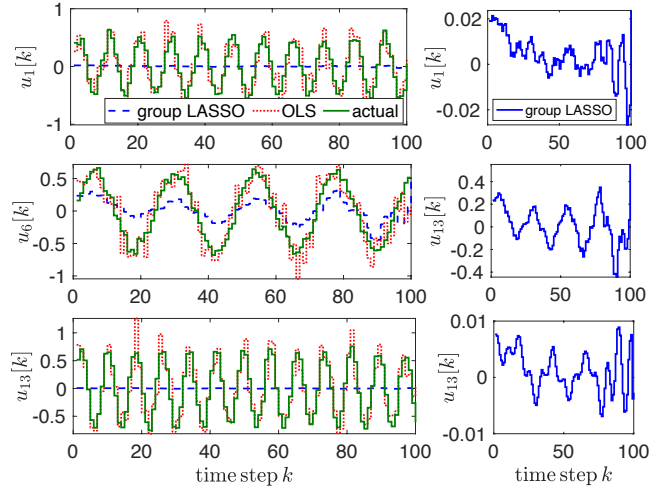


Fig. 4. FO inputs recovered by the group LASSO and OLS estimators. We used [30] to compute the OLS estimate using the locations recovered by the group LASSO. (Left panel) As predicted by Theorem 6 the OLS provides a better estimate than the LASSO estimator. (Right panel) zoomed plot of the group LASSO estimate.

estimated) inputs also increase with N . Given the relation in (36), the estimation error of \mathbf{x}_0^* is slightly higher than that of the unknown input. Finally, for greater estimation accuracy, one can always use the reduced model-based OLS estimator.

In Fig. 6 we show the average mutual incoherence (MIC) in (25) as a function of p , for two cases: $\mathbf{x}_0 = \mathbf{0}$ and $\mathbf{x}_0 \neq \mathbf{0}$. We computed both ℓ_1 - and ℓ_2 -norm based MICs. As pointed out in Section IV and confirmed by our plots in the left panel of Fig. 6 ℓ_2 -norm based MIC assumption is stronger than the ℓ_1 -norm. Further, when $\mathbf{x}_0 = \mathbf{0}$, the MIC is satisfied (that is, less than one) for as few as $p = 6$ sensors. Here, $p = m^* + 1$. Instead, when $\mathbf{x}_0 \neq \mathbf{0}$, we need at least $p = 18$ sensors to ensure that MIC is below one. Given m^* , theoretical relationships between p and MIC is left for future research.

VI. CONCLUSIONS

We study a group LASSO estimator for locating the sources of unknown forced inputs as well as estimating these inputs along with the initial state from noisy measurements. We

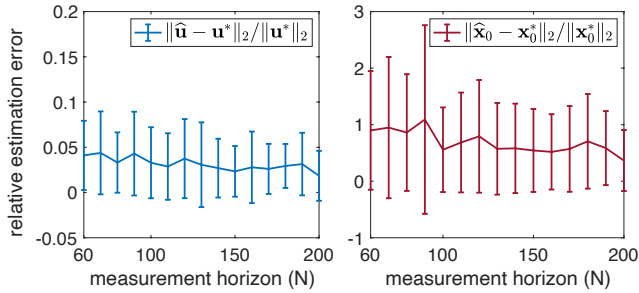


Fig. 5. Relative estimation error. Left panel: unknown inputs. Right panel: initial state.

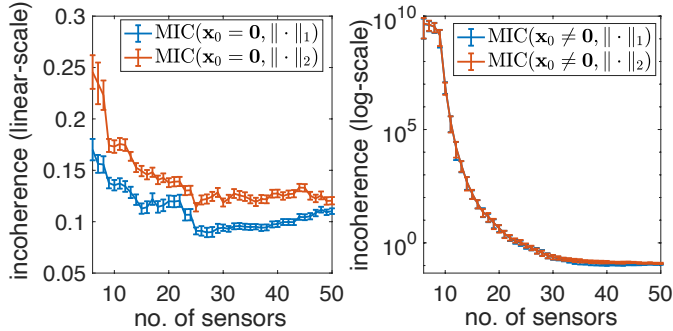


Fig. 6. Mutual incoherence vs. number of sensors. Left: $\mathbf{x}_0^* = \mathbf{0}$. Right: $\mathbf{x}_0^* \neq \mathbf{0}$

derive sufficient conditions under which the group LASSO estimate is location- and estimation-recovery consistent. In doing so, we have extended the existing theory of the group LASSO estimator for static regression models to linear models generated by d -delay (left) invertible linear dynamical systems with unknown initial state. Our results establish a relationship between time- and frequency-domain mutual incoherence conditions. The latter condition provides insight into the structural aspects of transfer matrices associated with the zero and non-zero inputs. Finally, we have validated the performance of our proposed group LASSO estimator via simulations.

Possible future work includes extending the group LASSO framework for linear and non-linear systems with additive state noise. As pointed out in Section IV-A in the presence of state noise, whitening the measurements can be detrimental to the performance of group LASSO. Two directions seem promising to tackle this issue: (i) to use the puffer-transformation method in [34] and (ii) to consider the group LASSO estimator for predictor or innovation form of the state-space model.

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VII. APPENDIX

A. KKT conditions and PDW Construction

Proposition 9. (Karush-Kuhn-Tucker (KKT) conditions) A necessary and sufficient condition for $(\hat{\mathbf{x}}_0, \hat{\mathbf{u}})$, with $\hat{\mathbf{u}}^T = [\hat{\mathbf{u}}_1^T, \dots, \hat{\mathbf{u}}_{m^*}^T]$, to be a solution of (11) is

$$-\frac{1}{T} \mathbf{O}^T [\mathbf{y} - \mathbf{O}\hat{\mathbf{x}}_0 - \sum_{j=1}^m \mathbf{J}_j \hat{\mathbf{u}}_j] = \mathbf{0} \quad (37)$$

$$-\frac{1}{T} \mathbf{J}_i^T [\mathbf{y} - \mathbf{O}\hat{\mathbf{x}}_0 - \sum_{j=1}^m \mathbf{J}_j \hat{\mathbf{u}}_j] + \lambda_T \hat{\mathbf{z}}_j = \mathbf{0} \quad (38)$$

for all $j \in \{1, \dots, m\}$. Here, $\hat{\mathbf{z}}_j$ is the subgradient of $\|\hat{\mathbf{u}}_j\|_2$; that is, $\hat{\mathbf{z}}_j = \hat{\mathbf{u}}_j / \|\hat{\mathbf{u}}_j\|_2$ if $\hat{\mathbf{u}}_j \neq \mathbf{0}$, else $\hat{\mathbf{z}}_j \in \{\mathbf{q} : \|\mathbf{q}\|_2 \leq 1\}$.

The proof is given in [31]. Without loss of generality let $S = \{1, \dots, m^*\}$ and $S^c = \{m^* + 1, \dots, m\}$. Let $\hat{\mathbf{u}}_S^T[k] = [\hat{\mathbf{u}}_1^T[k], \dots, \hat{\mathbf{u}}_{m^*}^T[k]]$, for all $k \in \{0, \dots, N\}$, where $\hat{\mathbf{u}}_j[k]$ is the k -th entry of $\hat{\mathbf{u}}_j$. Define $\hat{\mathbf{u}}_S^T = [\hat{\mathbf{u}}_S^T[0], \dots, \hat{\mathbf{u}}_S^T[N]]$. Thus,

$$[\hat{\mathbf{u}}_1^T, \dots, \hat{\mathbf{u}}_{m^*}^T]^T = \mathbf{P} \hat{\mathbf{u}}_S, \quad (39)$$

for some permutation matrix \mathbf{P} . Further, we can verify that $[\mathbf{J}_1, \dots, \mathbf{J}_{m^*}] \mathbf{P} = \mathbf{J}_S$ (as in (13)). Let $\hat{\beta}_S \triangleq [\hat{\mathbf{x}}_0^T, \hat{\mathbf{u}}_S^T]^T$. Then,

$$\mathbf{O}\hat{\mathbf{x}}_0 + \sum_{j \in S} \mathbf{J}_j \hat{\mathbf{u}}_j = \underbrace{[\mathbf{O} \quad \mathbf{J}_S]}_{\Psi_S} \hat{\beta}_S \quad (40)$$

Using these facts, expressions in (37)-(38) can be written as

$$-\frac{1}{T} \begin{bmatrix} \Psi_S^T \\ \mathbf{J}_{S^c}^T \end{bmatrix} [\mathbf{y} - \mathbf{O}\hat{\mathbf{x}}_0 - \sum_{j=1}^m \mathbf{J}_j \hat{\mathbf{u}}_j] + \lambda_T \begin{bmatrix} \mathbf{0} \\ \hat{\mathbf{z}}_{S^c} \end{bmatrix} = \begin{bmatrix} \mathbf{0} \\ \mathbf{0} \end{bmatrix}, \quad (41)$$

where $\tilde{\mathbf{J}}_{S^c} = [\mathbf{J}_{m^*+1}, \dots, \mathbf{J}_m]$, $\hat{\mathbf{z}}_S^T = [\hat{\mathbf{z}}_1^T, \dots, \hat{\mathbf{z}}_{m^*}^T]$, and $\hat{\mathbf{z}}_{S^c}^T = [\hat{\mathbf{z}}_{m^*+1}^T, \dots, \hat{\mathbf{z}}_m^T]$.

Primal-dual witness (PDW) construction: We prove Theorems 6 and 4 using the PDW construction technique [4] [29]:

- Set $\hat{\mathbf{u}}_j = \mathbf{0}$, for all $j \in S^c$.
- Let $(\hat{\mathbf{x}}_0, \hat{\mathbf{u}}_1, \dots, \hat{\mathbf{u}}_{m^*})$ be the solution of the sub-problem:

$$\min_{\substack{\hat{\mathbf{x}}_0; \\ \mathbf{u}_1, \dots, \mathbf{u}_{m^*}}} \frac{1}{2T} \left\| \mathbf{y} - \mathbf{O}\hat{\mathbf{x}}_0 - \sum_{j=1}^{m^*} \mathbf{J}_j \mathbf{u}_j \right\|_2^2 + \lambda_T \sum_{j=1}^{m^*} \|\mathbf{u}_j\|_2. \quad (42)$$

⁴The PDW construction is not an algorithm for solving the group LASSO problem in (11): This is because to solve the sub-problem in step (b) of PDW, we need to know the active set S . However, PDW construction technique helps to prove consistency results for the LASSO type problems.

Choose the sub-gradient $\hat{\mathbf{z}}_S = [\hat{\mathbf{z}}_1^T, \dots, \hat{\mathbf{z}}_{m^*}^T]^T$ such that

$$-\frac{1}{T} \Psi_S^T [\mathbf{y} - \mathbf{O}\hat{\mathbf{x}}_0 - \sum_{j=1}^{m^*} \mathbf{J}_j \hat{\mathbf{u}}_j] + \lambda_T \begin{bmatrix} \mathbf{0} \\ \mathbf{P}^T \hat{\mathbf{z}}_S \end{bmatrix} = \mathbf{0}. \quad (43)$$

- Solve $\hat{\mathbf{z}}_{S^c} = [\hat{\mathbf{z}}_{m^*+1}^T, \dots, \hat{\mathbf{z}}_m^T]^T$ using (41), and check if $\|\hat{\mathbf{z}}_j\|_2 \leq 1$, for all $j \in S^c = \{m^* + 1, \dots, m\}$.

By construction, $(\hat{\mathbf{x}}_0, \hat{\mathbf{u}}_1, \dots, \hat{\mathbf{u}}_{m^*})$, $\hat{\mathbf{z}}_S$, and $\hat{\mathbf{z}}_{S^c}$ that we determined in steps (a), and (b) satisfy conditions in (41). The PDW construction is said to be successful if $\hat{\mathbf{z}}_{S^c}$ satisfies the strict dual feasibility condition: $\|\hat{\mathbf{z}}_j\|_2 \leq 1$, for all $j \in S^c$.

B. Proofs of Theorems in Section IV

For the estimate in (42), define

$$\hat{\beta}_{\text{PDW}} = (\hat{\mathbf{x}}_0, \hat{\mathbf{u}}_1, \dots, \hat{\mathbf{u}}_{m^*}, \underbrace{\mathbf{0}_{(N+1)}, \dots, \mathbf{0}_{(N+1)}}_{m-m^*}). \quad (44)$$

Lemma 10. Suppose that the PDW construction succeeds. If delay $d > 0$, $\hat{\beta} = \hat{\beta}_{\text{PDW}}$ is an optimal solution of (11). If $d = 0$, $\hat{\beta} = \hat{\beta}_{\text{PDW}}$ is the "unique" optimal solution.

Proof. We follow the proof technique in [29] Lemma 7.23]. Let $d \geq 0$. Because the PDW construction succeeds, $\hat{\beta}_{\text{PDW}}$ is an optimal solution of (11) with subgradient vector $\hat{\mathbf{z}}^T = [\hat{\mathbf{z}}_0^T, \hat{\mathbf{z}}_1^T, \dots, \hat{\mathbf{z}}_m^T]$ satisfying $\hat{\mathbf{z}}_0 = \mathbf{0}$, $\|\hat{\mathbf{z}}_j\|_2 = 1$ for $j \in S$, and $\|\hat{\mathbf{z}}_j\|_2 < 1$ for $j \in S^c$. We now show that any optimal solution of (11) is supported on the set S .

With a slight abuse of notation, let $\mathbf{u}^T = [\mathbf{x}_0^T, \mathbf{u}_1^T, \dots, \mathbf{u}_m^T]$ and denote $F(\mathbf{u}) = \frac{1}{2T} \|\mathbf{y} - \mathbf{O}\mathbf{x}_0 + \sum_{j=1}^m \mathbf{J}_j \mathbf{u}_j\|_2^2$. Let $\nabla F(\mathbf{u})$ be the gradient of $F(\mathbf{u})$ with respect to \mathbf{u} . Then, for any other optimal solution $\tilde{\mathbf{u}}$ of (11), we have $F(\tilde{\mathbf{u}}) + \lambda_T \hat{\mathbf{z}}^T \tilde{\mathbf{u}} = F(\hat{\mathbf{u}}) + \lambda_T \sum_{j=1}^m \|\tilde{\mathbf{u}}_j\|_2$. The last equality follows because $\sum_{j=1}^m \hat{\mathbf{z}}_j^T \tilde{\mathbf{u}}_j = \sum_{j=1}^m \|\tilde{\mathbf{u}}_j\|_2$. Hence, $F(\hat{\mathbf{u}}) - \lambda_T \hat{\mathbf{z}}^T (\tilde{\mathbf{u}} - \hat{\mathbf{u}}) = F(\tilde{\mathbf{u}}) + \lambda_T \sum_{j=1}^m \|\tilde{\mathbf{u}}_j\|_2 - \lambda_T \hat{\mathbf{z}}^T \tilde{\mathbf{u}}$. Instead, from the zero-subgradient conditions in (37)-(38), we have $\lambda_T \hat{\mathbf{z}} = -\nabla F(\hat{\mathbf{u}})$. Putting the pieces together, we have

$$F(\hat{\mathbf{u}}) + \nabla F(\hat{\mathbf{u}})^T (\tilde{\mathbf{u}} - \hat{\mathbf{u}}) - F(\tilde{\mathbf{u}}) = \lambda_T \left(\sum_{j=1}^m \|\tilde{\mathbf{u}}_j\|_2 - \hat{\mathbf{z}}^T \tilde{\mathbf{u}} \right).$$

By convexity of F , the left-hand side is negative. As a result, $\sum_{j=1}^m \|\tilde{\mathbf{u}}_j\|_2 \leq \hat{\mathbf{z}}^T \tilde{\mathbf{u}} = \sum_{j=1}^m \hat{\mathbf{z}}_j^T \tilde{\mathbf{u}}_j$, where $\hat{\mathbf{z}}_0 = \mathbf{0}$. Since we also have $\sum_{j=1}^m \hat{\mathbf{z}}_j^T \tilde{\mathbf{u}}_j \leq \sum_{j=1}^m \|\hat{\mathbf{z}}_j\|_2 \|\tilde{\mathbf{u}}_j\|_2 \leq \sum_{j=1}^m \|\tilde{\mathbf{u}}_j\|_2$, we must have $\sum_{j=1}^m \|\tilde{\mathbf{u}}_j\|_2 = \sum_{j=1}^m \hat{\mathbf{z}}_j^T \tilde{\mathbf{u}}_j$. Because $\|\hat{\mathbf{z}}_j\|_2 < 1$ for $j \in S^c$, the above equality can only occur if $\tilde{\mathbf{u}}_j = \mathbf{0}$, for all $j \in S^c$. To see this notice that $\sum_{j=1}^m \hat{\mathbf{z}}_j^T \tilde{\mathbf{u}}_j = \sum_{j \in S} \hat{\mathbf{z}}_j^T \tilde{\mathbf{u}}_j + \sum_{j \in S^c} \|\hat{\mathbf{z}}_j\|_2 \|\tilde{\mathbf{u}}_j\|_2 \cos(\theta_j)$, where θ_j is the angle between $\hat{\mathbf{z}}_j$ and $\tilde{\mathbf{u}}_j$, and $\|\hat{\mathbf{z}}_j\|_2 \cos(\theta_j) \in (-1, 1)$. Thus, all optimal solutions $\hat{\beta}$ are such that $\hat{\beta}_j = \mathbf{0}$ for all $j \in S^c$. These solutions can be obtained by solving (42). Finally, for $d = 0$, the assumption in (A2) ensures that (42) is strictly convex, and hence, $\hat{\beta}$ is a unique minimizer. \square

Proof of Theorem 4. Suppose the PDW construction succeeds. The proof of part (a) is given in Lemma 10. Further, in view of Lemma 10, $\hat{\beta} = \hat{\beta}_{\text{PDW}}$ is an optimal solution of (11). Thus, all the optimal input vectors are supported on the set S , i.e., $\hat{S} \subset S$, where $\hat{S} = \{j : \hat{\mathbf{u}}_j \neq \mathbf{0}\}$; thus, part (b) holds.

We show that the PDW construction succeeds with probability at least $1 - 2 \exp(-T\delta^2/2)$ by showing that $\|\hat{\mathbf{z}}_j\|_2 \leq 1$,

for all $j \in S^c$. Here, $\hat{\mathbf{z}}_j$ is determined in the step (c) of PDW construction. Let $\hat{\boldsymbol{\beta}}_S$ be as in (40). By substituting \mathbf{y} (given in (12)) and $\hat{\mathbf{u}}_{S^c} = \mathbf{0}$ in (41), we obtain

$$\frac{1}{T} \begin{bmatrix} \boldsymbol{\Psi}_S^T \boldsymbol{\Psi}_S & \boldsymbol{\Psi}_S^T \tilde{\mathbf{J}}_{S^c} \\ \tilde{\mathbf{J}}_{S^c}^T \boldsymbol{\Psi}_S & \tilde{\mathbf{J}}_{S^c}^T \mathbf{J}_{S^c} \end{bmatrix} \begin{bmatrix} \boldsymbol{\beta}_S^* - \hat{\boldsymbol{\beta}}_S \\ \mathbf{0} \end{bmatrix} + \frac{1}{T} \begin{bmatrix} \boldsymbol{\Psi}_S^T \\ \tilde{\mathbf{J}}_{S^c}^T \end{bmatrix} \mathbf{v} = \lambda_T \begin{bmatrix} \mathbf{0} \\ \mathbf{P}^T \hat{\mathbf{z}}_{S^c} \\ \hat{\mathbf{z}}_{S^c} \end{bmatrix}. \quad (45)$$

Using the second block equation of (45), solve for $\hat{\mathbf{z}}_{S^c}$ as

$$\hat{\mathbf{z}}_{S^c} = \tilde{\mathbf{J}}_{S^c}^T \boldsymbol{\Psi}_S \left[\frac{\boldsymbol{\Psi}_S^T \boldsymbol{\Psi}_S}{\lambda_T T} (\boldsymbol{\beta}_S^* - \hat{\boldsymbol{\beta}}_S) \right] + \tilde{\mathbf{J}}_{S^c}^T \left(\frac{\mathbf{v}}{\lambda_T T} \right), \quad (46)$$

where we used the fact $\boldsymbol{\Psi}_S = \boldsymbol{\Psi}_S \boldsymbol{\Psi}_S^\dagger \boldsymbol{\Psi}_S$. On the other hand, from the top block equation in (45), we have

$$\frac{1}{T} \boldsymbol{\Psi}_S^T \boldsymbol{\Psi}_S (\boldsymbol{\beta}_S^* - \hat{\boldsymbol{\beta}}_S) + \frac{1}{T} \boldsymbol{\Psi}_S^T \mathbf{v} = \lambda_T \begin{bmatrix} \mathbf{0} \\ \mathbf{P}^T \hat{\mathbf{z}}_S \end{bmatrix}. \quad (47)$$

Pre-multiply both sides of the equality in (47) with $(\boldsymbol{\Psi}_S^T \boldsymbol{\Psi}_S)^\dagger$. Then, use the identities $(\boldsymbol{\Psi}_S^T \boldsymbol{\Psi}_S)^\dagger (\boldsymbol{\Psi}_S^T \boldsymbol{\Psi}_S)^\dagger = \boldsymbol{\Psi}_S^\dagger \boldsymbol{\Psi}_S$ and $\boldsymbol{\Psi}_S^\dagger = (\boldsymbol{\Psi}_S^T \boldsymbol{\Psi}_S)^\dagger \boldsymbol{\Psi}_S^T$ (see [35]) to get the following:

$$\boldsymbol{\Psi}_S^\dagger \boldsymbol{\Psi}_S (\boldsymbol{\beta}_S^* - \hat{\boldsymbol{\beta}}_S) = -\boldsymbol{\Psi}_S^\dagger \mathbf{v} + T \lambda_T (\boldsymbol{\Psi}_S^T \boldsymbol{\Psi}_S)^\dagger \begin{bmatrix} \mathbf{0} \\ \mathbf{P}^T \hat{\mathbf{z}}_S \end{bmatrix}. \quad (48)$$

Let $\boldsymbol{\Gamma}_S = [\mathbf{I} - (\boldsymbol{\Psi}_S \boldsymbol{\Psi}_S^\dagger)]$. By substituting (48) in the first term of the second equality in (46), we can simplify $\hat{\mathbf{z}}_{S^c}$ as

$$\hat{\mathbf{z}}_{S^c} = \tilde{\mathbf{J}}_{S^c}^T (\boldsymbol{\Psi}_S^\dagger)^T \begin{bmatrix} \mathbf{0} \\ \mathbf{P}^T \hat{\mathbf{z}}_S \end{bmatrix} + \tilde{\mathbf{J}}_{S^c}^T \boldsymbol{\Gamma}_S \left(\frac{\mathbf{v}}{\lambda_T T} \right), \quad (49)$$

where we used the fact $(\boldsymbol{\Psi}_S^\dagger)^T = \boldsymbol{\Psi}_S (\boldsymbol{\Psi}_S^T \boldsymbol{\Psi}_S)^\dagger$. Thus,

$$\hat{\mathbf{z}}_j = \mathbf{J}_j^T (\boldsymbol{\Psi}_S^\dagger)^T \begin{bmatrix} \mathbf{0} \\ \mathbf{P}^T \hat{\mathbf{z}}_S \end{bmatrix} + \mathbf{J}_j^T \boldsymbol{\Gamma}_S \left(\frac{\mathbf{v}}{\lambda_T T} \right), \quad \forall j \in S^c \quad (50)$$

By the sub-multiplicative property of norms, for any $j \in S^c$,

$$\begin{aligned} \left\| \mathbf{J}_j^T (\boldsymbol{\Psi}_S^\dagger)^T \begin{bmatrix} \mathbf{0} \\ \mathbf{P}^T \hat{\mathbf{z}}_S \end{bmatrix} \right\|_2 &\leq \max_{j \in S^c} \|\mathbf{J}_j^T (\boldsymbol{\Psi}_S^\dagger)^T\|_2 \left\| \begin{bmatrix} \mathbf{0} \\ \mathbf{P}^T \hat{\mathbf{z}}_S \end{bmatrix} \right\|_2 \\ &\leq \frac{\alpha}{m^*} \|\mathbf{P}^T \hat{\mathbf{z}}_S\|_2 \leq \frac{\alpha}{m^*} \sum_{j \in S} \|\hat{\mathbf{z}}_j\|_2 \leq \alpha. \end{aligned}$$

where $\alpha \leq 1$ is given in (25) and we used the fact that $\|\hat{\mathbf{z}}_j\|_2 \leq 1$ (see Proposition 9), for $j \in S$, and $\|\mathbf{P}^T\|_2 \leq 1$. As a result, from (50) and the preceding inequality, we have

$$\max_{j \in S^c} \|\hat{\mathbf{z}}_j\|_2 \leq \alpha + \max_{j \in S^c} \left\| \mathbf{J}_j^T \boldsymbol{\Gamma}_S \left(\frac{\mathbf{v}}{\lambda_T T} \right) \right\|_2 = \alpha + \kappa_2. \quad (51)$$

On the other hand, in light of Lemma 11, $\kappa_2 \leq 0.5(1 - \alpha)$ with probability at least $1 - 2 \exp(-\delta^2 T/2)$, for $\delta > 0$. Putting together the pieces, we have $\max_{j \in S^c} \|\hat{\mathbf{z}}_j\|_2 \leq 0.5(1 + \alpha) < 1$, thereby establishing the strict dual feasibility condition.

Part (c): From Assumption (A2) and Proposition 1 we have

$$\boldsymbol{\Psi}_S^\dagger \boldsymbol{\Psi}_S = \text{Blkdiag}(\mathbf{I}_n, \mathbf{I}_{t_S}, \boldsymbol{\Psi}_{S,[d-1:0]}^\dagger \boldsymbol{\Psi}_{S,[d-1:0]}), \quad (52)$$

where $t_S = (N - d + 1)m^*$. Thus, we have

$$\mathbf{u}_{S,[0:N-d]}^* - \hat{\mathbf{u}}_{S,[0:N-d]} = \boldsymbol{\Pi}_{S,[0:N-d]} \boldsymbol{\Psi}_S^\dagger \boldsymbol{\Psi}_S (\boldsymbol{\beta}_S^* - \hat{\boldsymbol{\beta}}_S), \quad (53)$$

where $\boldsymbol{\Pi}_{S,[0:N-d]} = [\mathbf{0}_{t_S \times n} \mathbf{I}_{t_S \times t_S} \mathbf{0}_{t_S \times dm^*}]$ and

$$\boldsymbol{\beta}_S^* - \hat{\boldsymbol{\beta}}_S = \begin{bmatrix} \mathbf{x}_0^* - \hat{\mathbf{x}}_0 \\ \mathbf{u}_{S,[0:N-d]}^* - \hat{\mathbf{u}}_{S,[0:N-d]} \\ \mathbf{u}_{S,[N-d+1:0]}^* - \hat{\mathbf{u}}_{S,[N-d+1:0]} \end{bmatrix}. \quad (54)$$

For brevity, let $\boldsymbol{\Pi} = \boldsymbol{\Pi}_{S,[0:N-d]}$. From (53) and (48), and the sub-multiplicative property of norms, we have

$$\begin{aligned} \|\mathbf{u}_{S,[0:N-d]}^* - \hat{\mathbf{u}}_{S,[0:N-d]}\|_\infty &\leq \|\boldsymbol{\Pi} \boldsymbol{\Psi}_S^\dagger \mathbf{v}\|_\infty \\ &\quad + \lambda_T \left\| \boldsymbol{\Pi} (\boldsymbol{\Psi}_S^T \boldsymbol{\Psi}_S / T)^\dagger \right\|_\infty, \end{aligned} \quad (55)$$

where we used the fact $\|\hat{\mathbf{z}}_S\|_\infty \leq 1$. The second term is deterministic. Instead, the first term is random, and, from Lemma 3, it is upper bounded by $\sigma / \sqrt{c_{\min}(\sqrt{2 \log(t_S)/T} + \delta)}$ with probability at least $1 - 2 \exp(-T\delta^2/2)$. Finally, the left-hand side of (55) can be written as $\max_{j \in S} \|\mathbf{u}_{j,[0:N-d]}^* - \hat{\mathbf{u}}_{j,[0:N-d]}\|_\infty$. Putting the pieces together, we have the inequality in (27).

Part (d): By the triangle inequality, for all $j \in S$, we have

$$\begin{aligned} \|\mathbf{u}_{j,[0:N-d]}^*\|_\infty &= \|\mathbf{u}_{j,[0:N-d]}^* - \hat{\mathbf{u}}_{j,[0:N-d]} + \hat{\mathbf{u}}_{j,[0:N-d]}\|_\infty \\ &\leq \|\mathbf{u}_{j,[0:N-d]}^* - \hat{\mathbf{u}}_{j,[0:N-d]}\|_\infty + \|\hat{\mathbf{u}}_{j,[0:N-d]}\|_\infty \\ &\stackrel{(i)}{\leq} \beta_{\min} + \|\hat{\mathbf{u}}_{j,[0:N-d]}\|_\infty, \end{aligned}$$

where (i) follows from part (c). Thus, $\|\hat{\mathbf{u}}_{j,[0:N-d]}\|_\infty > 0$ if $\|\mathbf{u}_{j,[0:N-d]}^*\|_\infty > \beta_{\min}$. This observation together with $\hat{S} \subseteq S$ in part (a) implies that $\hat{S} = S$.

Finally, the probability stated in the theorem is obtained by taking the union bound of the event where the dual feasibility holds and the event where ℓ_∞ bounds hold. \square

Proof of Theorem 7: Consider the auxiliary system $\mathbf{x}[k+1] = \mathbf{A}\mathbf{x}[k] + \mathbf{b}_j \mathbf{u}_j^*[k]$, where $j \in S^c$ and $u_j^*[k] = 0$, $k \geq N$. Let $\mathbf{x}[0] = \mathbf{0}$. Thus, $\mathbf{y} = \mathbf{J}_j \mathbf{u}_j^*$, where \mathbf{J}_j is given by (7) and \mathbf{y} and \mathbf{u}_j^* as in (6). Let $\boldsymbol{\Psi}_S$ be as in (12), and consider

$$\begin{bmatrix} \tilde{\mathbf{y}}[0] \\ \vdots \\ \tilde{\mathbf{y}}[N] \end{bmatrix} \triangleq \boldsymbol{\Psi}_S^\dagger \begin{bmatrix} \mathbf{y}[0] \\ \vdots \\ \mathbf{y}[N] \end{bmatrix} = \boldsymbol{\Psi}_S^\dagger \mathbf{J}_j \mathbf{u}_j^* = \boldsymbol{\Psi}_S^\dagger \mathbf{J}_j \begin{bmatrix} u_j^*[0] \\ \vdots \\ u_j^*[N] \end{bmatrix}, \quad (56)$$

By assumption we have $\text{nRank} \mathcal{Z}_S = n + m^*$. Thus, for all $z \notin \text{spec}(A)$, $\mathcal{G}_S[z]$ has full column rank and $\mathcal{G}_S^+[z] = [\mathcal{G}_S[z]^T \mathcal{G}_S[z]]^{-1} \mathcal{G}_S[z]^T$ and $\mathcal{G}_S^+[z] \mathcal{G}_S[z] = z^{-d} \mathbf{I}$; see [8] Theorem 1. Let $\tilde{\mathbf{y}}[z]$ be the \mathcal{Z} -transform of $\{\tilde{\mathbf{y}}[k]\}_{k=0}^\infty$. Then by using the construction given in [36] pp. 49-50] and the uniqueness of pseudo inverse [26], we have $\tilde{\mathbf{y}}[z] = z^{-d} \mathcal{H}[z] \mathbf{u}_j^*[z]$, where $\mathcal{H}_j[z] = \mathcal{G}_S^+[z] \mathcal{G}_j[z]$, for all $z \notin \text{spec}(A)$.

From Parseval's theorem, we have the following bound:

$$\begin{aligned} \sqrt{\sum_{k=0}^\infty \|\tilde{\mathbf{y}}[k]\|_2^2} &= \sqrt{\frac{1}{2\pi} \int_{-\pi}^\pi \|\tilde{\mathbf{y}}[e^{j\omega}]\|_2^2 d\omega} \\ &= \sqrt{\frac{1}{2\pi} \int_{-\pi}^\pi \|e^{-dj\omega} \mathcal{H}_j[e^{j\omega}] \mathbf{u}_j^*[e^{j\omega}]\|_2^2 d\omega} \\ &\leq \sup_{\{\omega \in [-\pi, \pi]\}} \|\mathcal{H}_j[e^{j\omega}]\|_2 \sqrt{\frac{1}{2\pi} \int_{-\pi}^\pi \|\mathbf{u}_j^*[e^{j\omega}]\|_2^2 d\omega} \\ &= \sup_{\{z \in \mathbb{C}: |z|=1\}} \|\mathcal{H}_j[z]\|_2 \|\mathbf{u}_j^*\|_2. \end{aligned} \quad (57)$$

For the final inequality, we once again used Parseval's theorem and the fact that $u^*[k] = 0$, for all $k > N$.

On the other hand, from (56) and (57), we have

$$\begin{aligned} \|\Psi_S^\dagger \mathbf{J}_j\|_2 &= \sup_{\|\mathbf{u}_j^*\|_2=1} \|\Psi_S^\dagger \mathbf{J}_j \mathbf{u}_j^*\|_2 = \sup_{\|\mathbf{u}_j^*\|_2=1} \sqrt{\sum_{k=0}^N \|\tilde{\mathbf{y}}[k]\|_2^2} \\ &\leq \sup_{\|\mathbf{u}_j^*\|_2=1} \sqrt{\sum_{k=0}^{\infty} \|\tilde{\mathbf{y}}[k]\|_2^2} \leq \sup_{\{z \in \mathbb{C}: |z|=1\}} \|\mathcal{H}_j[z]\|_2. \end{aligned}$$

Thus $\max_{\{j \in S^c\}} \sup_{\{z \in \mathbb{C}: |z|=1\}} \|\mathcal{H}[z]\|_2 \leq \alpha/m^*$ implies that $\|\Psi_S^\dagger \mathbf{J}_j\|_2 \leq \alpha/m^*$. The proof is now complete. \square

Proof of Theorem (6): From Theorem 4, $S = \hat{S}$ holds with probability at least $1 - 4 \exp(-T\delta^2/2)$. Thus, from (30), with the same probability, we have

$$\hat{\beta}_{\hat{S},[0:N-d]}^{(OLS)} = \hat{\beta}_{S,[0:N-d]}^{(OLS)} = \tilde{\Pi}_{S,[0:N-d]} \Psi^\dagger \mathbf{y}, \quad (58)$$

where $\tilde{\Pi}_S \triangleq \tilde{\Pi}_{S,[0:N-d]} = [\mathbf{I}_{n+t_S} \mathbf{0}_{(n+t_S) \times dm^*}]$.

Since $\mathbf{y} \sim \mathcal{N}(\beta_{S,[0:N-d]}^*, \sigma^2 \mathbf{I})$, it follows that

$$\hat{\beta}_{S,[0:N-d]}^{(OLS)} - \beta_{S,[0:N-d]}^* \sim \mathcal{N}(\mathbf{0}, \underbrace{\sigma^2 \tilde{\Pi}_S (\Psi_S^\top \Psi_S)^\dagger \tilde{\Pi}_S^\top}_{\triangleq \Sigma \in \mathbb{R}^{n+t_S \times n+t_S}}). \quad (59)$$

We now upper bound $\|\Sigma\|_2$. Recall from (19) and Proposition 1 (ii) that $\Psi_S = [\Psi_{S,[N:d]} \Psi_{S,[d-1:0]}]$ and $\mathcal{R}(\Psi_{S,[N:d]}) \cap \mathcal{R}(\Psi_{S,[d-1:0]}) = \{0\}$. Let $\mathbf{M} = [\mathbf{I} - \Psi_{S,[d-1:0]} \Psi_{S,[d-1:0]}^\dagger]$. Then, by invoking [26] Lemma D], we have

$$\tilde{\Pi}_S (\Psi_S^\top \Psi_S)^\dagger \tilde{\Pi}_S^\top = [(\mathbf{M} \Psi_{S,[N:d]})^\dagger (\mathbf{M} \Psi_{S,[N:d]})]^\top. \quad (60)$$

Since $\mathbf{M}^* = \mathbf{M} = \mathbf{M}^2$, from Assumption (A2), it follows that $\|\Sigma\|_2 \leq \sigma^2/(Tc_{\min})$.

From the second concentration result in Lemma 13 we have

$$\begin{aligned} \left\| \beta_{S,[0:N-d]}^* - \hat{\beta}_{S,[0:N-d]}^{(OLS)} \right\|_2 &\leq \frac{4\sigma}{\sqrt{c_{\min}}} \left\{ \sqrt{\frac{(n+t_S)}{T}} \right\} \\ &+ \frac{2\sigma}{\sqrt{c_{\min}}} \left\{ \sqrt{\frac{1}{T} \log\left(\frac{1}{\delta_1}\right)} \right\}, \quad (61) \end{aligned}$$

with probability at least $1 - \delta_1$ for $\delta_1(0, 1)$. The statement of the theorem follows by taking an union bound over the events where (61) and (58) hold. \square

Lemma 11. *With the notation and assumptions stated in Theorem 4 we have $\mathbb{P}[\max_{j \in S^c} \|\mathbf{J}_j^\top \Gamma_S (\mathbf{v}/\lambda_T T)\|_2 \geq 0.5(1 - \alpha)] \leq 2 \exp(-T\delta^2/2)$, where $\Gamma_S = [\mathbf{I} - \Psi_S \Psi_S^\dagger]$ and $\alpha \in [0, 1)$.*

Proof. Let $\tilde{\alpha} = 0.5(1 - \alpha)$ and take the union bound to get

$$\begin{aligned} \mathbb{P} \left[\max_{j \in S^c} \left\| \mathbf{J}_j^\top \Gamma_S \left(\frac{\mathbf{v}}{\lambda_T T} \right) \right\|_2 \geq \tilde{\alpha} \right] \\ \leq \sum_{j \in S^c} \mathbb{P} \left[\left\| \mathbf{J}_j^\top \Gamma_S \left(\frac{\mathbf{v}}{\lambda_T T} \right) \right\|_2 \geq \tilde{\alpha} \right]. \quad (62) \end{aligned}$$

Since $\mathbf{v} \sim \mathcal{N}(\mathbf{0}, \sigma^2 \mathbf{I})$, we have $\mathbf{J}_j^\top \Gamma_S (\mathbf{v}/\lambda_T T) \sim \mathcal{N}(\mathbf{0}, \Sigma_j)$ with $\Sigma_j = \mathbf{J}_j^\top \Gamma_S \Gamma_S^\top \mathbf{J}_j / (\lambda_T^2 T^2)$. Furthermore, from the identity that $\|\mathbf{X}\mathbf{X}^\top\|_2 = \|\mathbf{X}\|_2^2$, the following is trivial.

$$\|\Sigma_j\|_2 = \frac{1}{\lambda_T^2 T^2} \|\mathbf{J}_j^\top \Gamma_S\|_2^2 \leq \frac{1}{\lambda_T^2 T^2} \|\mathbf{J}_j^\top\|_2^2 \leq \frac{C^2 T}{\lambda_T^2 T^2}. \quad (63)$$

The first inequality follows because Γ_S is a projection matrix and for the last inequality from the normalization assumption (A1). Invoking Lemma 13 we bound the inequality in (62) as

$$\sum_{j \in S^c} \mathbb{P} \left[\left\| \mathbf{J}_j^\top \Gamma_S \left(\frac{\mathbf{v}}{\lambda_T T} \right) \right\|_2 \geq \tilde{\alpha} \right] \leq \sum_{j \in S^c} c_N \exp \left(-\frac{\tilde{\alpha}^2 \lambda_T^2 T}{8\sigma^2 C^2} \right),$$

where $c_N = 5^{N+1}$. Since each term in the summand is the same, the right-hand side can be expressed as

$$\exp \left((N+1) \log(5) + \log(m - m_0) - \frac{\tilde{\alpha}^2 \lambda_T^2 T}{8\sigma^2 C^2} \right) \quad (64)$$

Substituting λ_T (see (26)) and $\tilde{\alpha} = 0.5(1 - \alpha)$ in (64), and simplifying it gives us the required bound. \square

Lemma 12. *With the notation and assumptions stated in Theorem 4 for $\delta \in [0, 1)$, we have $\mathbb{P}[\|\Pi_{S,[0:N-d]} \Psi_S^\dagger \mathbf{v}\|_\infty \geq \sigma/\sqrt{c_{\min}}(\sqrt{2 \log(t_S)/T} + \delta)] \leq 2 \exp(-T\delta^2/2)$.*

Proof. Recall that $\Pi_{S,[0:N-d]} = [\mathbf{0}_{t_S \times n} \mathbf{I}_{t_S \times t_S} \mathbf{0}_{t_S \times dm^*}]$ and $t_S = (N-d+1)m^*$. We drop the sub-script notation $[0 : N-d]$ in $\Pi_{S,[0:N-d]}$. Let $z_l = \mathbf{e}_l^\top \Pi_S \Psi_S^\dagger \mathbf{v}$ be the l^{th} entry of the vector $\Pi_S \Psi_S^\dagger \mathbf{v}$, where \mathbf{e}_l is the l^{th} canonical basis vector in \mathbb{R}^{t_S} . Thus $\|\Pi_{S,[0:N-d]} \Psi_S^\dagger \mathbf{v}\|_\infty = \max_{l \in 1 \dots t_S} |z_l|$ and for $\kappa \geq 0$ by invoking union bound we have

$$\mathbb{P} \left[\max_{l \in 1 \dots t_S} |z_l| \geq \kappa \right] \leq \sum_{l=1}^{t_S} \mathbb{P}[|z_l| \geq \kappa]. \quad (65)$$

We bound terms on the right-hand side by invoking standard concentration results. We first characterize the distribution of z_l . Since $\mathbf{v} \sim \mathcal{N}(\mathbf{0}, \sigma^2 \mathbf{I})$, we have $z_l \sim \mathcal{N}(0, \sigma_l^2)$, where

$$\begin{aligned} \sigma_l^2 &= \sigma^2 \mathbf{e}_l^\top \Pi_S \Psi_S^\dagger (\Pi_S \Psi_S^\dagger)^\top \mathbf{e}_l \leq \sigma^2 \lambda_{\max}(\Pi_S \Psi_S^\dagger (\Pi_S \Psi_S^\dagger)^\top) \\ &= \sigma^2 \|\Pi_S (\Psi_S^\top \Psi_S)^\dagger \Pi_S^\top\|_2 \\ &\leq \sigma^2 \|\tilde{\Pi}_S (\Psi_S^\top \Psi_S)^\dagger \tilde{\Pi}_S^\top\|_2 \\ &\leq \sigma^2/(Tc_{\min}). \quad (66) \end{aligned}$$

where $\tilde{\Pi}_S = [\mathbf{I}_{n+t_S} \mathbf{0}_{(n+t_S) \times dm^*}]$. The second inequality follows from interlacing property of singular values. The final inequality is showed in the proof of Theorem 6.

Since z_l is Gaussian, from [29] page 22] and (66), we have $\mathbb{P}[|z_l| \geq \kappa] \leq \exp(-\kappa^2/(2\sigma_l^2)) \leq \exp(-\kappa^2 T c_{\min}/(2\sigma^2))$. Substituting this inequality in (67), we find that

$$\mathbb{P} \left[\max_{l \in 1 \dots t_S} |z_l| \geq \kappa \right] \leq \exp \left(\log(t_S) - \frac{\kappa^2 T c_{\min}}{2\sigma^2} \right). \quad (67)$$

The result follows by letting $\kappa = \sigma/\sqrt{c_{\min}}(\sqrt{2 \log(t_S)/T} + \delta)$ and simplifying terms in the exponential term. \square

Lemma 13. *Let $\mathbf{p} \sim \mathcal{N}(\mathbf{0}, \Sigma)$, where $\Sigma \in \mathbb{R}^{l \times l}$ is a positive definite matrix. Then, $\mathbb{P}[\|\mathbf{p}\|_2 \geq t] \leq 5^l \exp(-t^2/(8\|\Sigma\|_2))$. Furthermore, $\|\mathbf{p}\|_2 \leq 4\sqrt{\|\Sigma\|_2} t + 2\sqrt{\|\Sigma\|_2 \log(1/\delta)}$ with probability at least $1 - \delta$ for $\delta(0, 1)$.*

Proof. Follows from Lemma 8.2 and Theorem 8.3 in [37]. \square